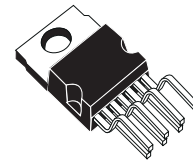


2.5A POWER SWITCHING REGULATOR

- 2.5A OUTPUT CURRENT
- 5.1V TO 40V OUTPUT VOLTAGE RANGE
- PRECISE ($\pm 2\%$) ON-CHIP REFERENCE
- HIGH SWITCHING FREQUENCY
- VERY HIGH EFFICIENCY (UP TO 90%)
- VERY FEW EXTERNAL COMPONENTS
- SOFT START
- INTERNAL LIMITING CURRENT
- THERMAL SHUTDOWN



HEPTAWATT

ORDERING NUMBERS: L4960 (Vertical)
L4960H (Horizontal)

DESCRIPTION

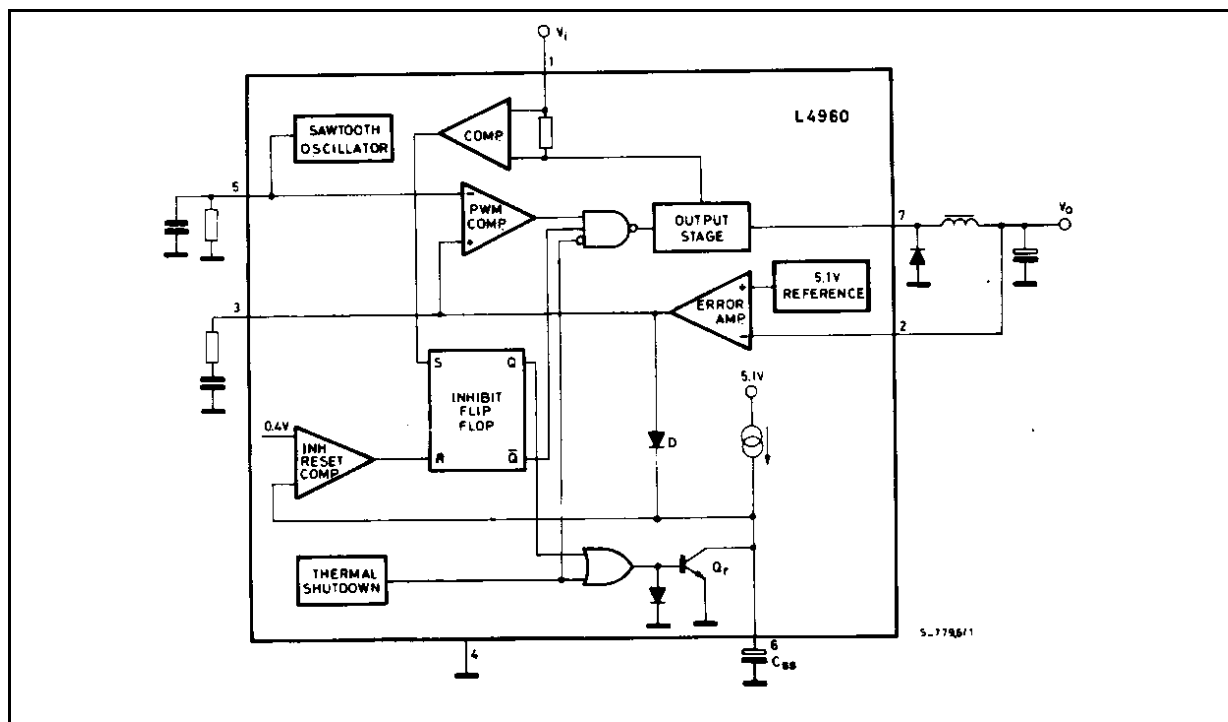
The L4960 is a monolithic power switching regulator delivering 2.5A at a voltage variable from 5V to 40V in step down configuration.

Features of the device include current limiting, soft start, thermal protection and 0 to 100% duty cycle for continuous operation mode.

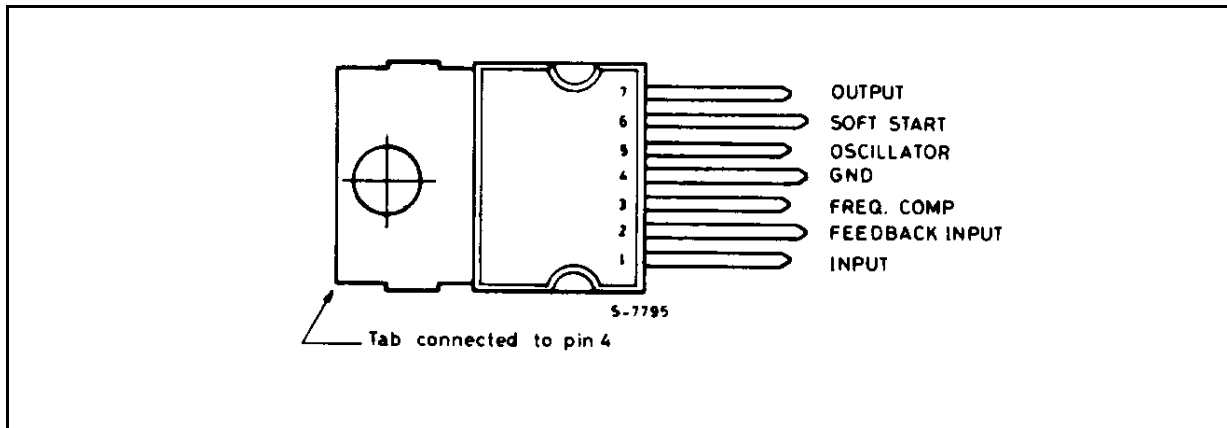
The L4960 is mounted in a Heptawatt plastic power package and requires very few external components.

Efficient operation at switching frequencies up to 150KHz allows a reduction in the size and cost of external filter components.

BLOCK DIAGRAM



PIN CONNECTION (Top view)



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_1	Input voltage	50	V
$V_1 - V_7$	Input to output voltage difference	50	V
V_7	Negative output DC voltage	-1	V
	Negative output peak voltage at $t = 0.1\mu s$; $f = 100KHz$	-5	V
V_3, V_6	Voltage at pin 3 and 6	5.5	V
V_2	Voltage at pin 2	7	V
I_3	Pin 3 sink current	1	mA
I_5	Pin 5 source current	20	mA
P_{tot}	Power dissipation at $T_{case} \leq 90^\circ C$	15	W
T_j, T_{stg}	Junction and storage temperature	-40 to 150	$^\circ C$

PIN FUNCTIONS

N°	NAME	FUNCTION
1	SUPPLY VOLTAGE	Unregulated voltage input. An internal regulator powers the internal logic.
2	FEEDBACK INPUT	The feedback terminal of the regulation loop. The output is connected directly to this terminal for 5.1V operation; it is connected via a divider for higher voltages.
3	FREQUENCY COMPENSATION	A series RC network connected between this terminal and ground determines the regulation loop gain characteristics.
4	GROUND	Common ground terminal.
5	OSCILLATOR	A parallel RC network connected to this terminal determines the switching frequency.
6	SOFT START	Soft start time constant. A capacitor is connected between this terminal and ground to define the soft start time constant. This capacitor also determines the average short circuit output current.
7	OUTPUT	Regulator output.

THERMAL DATA

Symbol	Parameter	Value	Unit
$R_{th\ j-case}$	Thermal resistance junction-case	max	4 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	50 °C/W

ELECTRICAL CHARACTERISTICS (Refer to the test circuit, $T_j = 25\text{ °C}$, $V_i = 35V$, unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
--------	-----------	-----------------	------	------	------	------

DYNAMIC CHARACTERISTICS

V_o	Output voltage range	$V_i = 46V$ $I_o = 1A$	V_{ref}		40	V
V_i	Input voltage range	$V_o = V_{ref}$ to 36V $I_o = 2.5A$	9		46	V
ΔV_o	Line regulation	$V_i = 10V$ to 40V $V_o = V_{ref}$ $I_o = 1A$		15	50	mV
ΔV_o	Load regulation	$V_o = V_{ref}$ $I_o = 0.5A$ to 2A		10	30	mV
V_{ref}	Internal reference voltage (pin 2)	$V_i = 9V$ to 46V $I_o = 1A$	5	5.1	5.2	V
$\frac{\Delta V_{ref}}{\Delta T}$	Average temperature coefficient of refer voltage	$T_j = 0\text{°C}$ to 125°C $I_o = 1A$		0.4		mV/°C
V_d	Dropout voltage	$I_o = 2A$		1.4	3	V
I_{om}	Maximum operating load current	$V_i = 9V$ to 46V $V_o = V_{ref}$ to 36V	2.5			A
I_{7L}	Current limiting threshold (pin 7)	$V_i = 9V$ to 46V $V_o = V_{ref}$ to 36V	3		4.5	A
I_{SH}	Input average current	$V_i = 46V$; output short-circuit		30	60	mA
η	Efficiency	$f = 100KHz$ $V_o = V_{ref}$		75		%
		$I_o = 2A$ $V_o = 12V$		85		%
SVR	Supply voltage ripple rejection	$\Delta V_i = 2V_{rms}$ fripple = 100Hz $V_o = V_{ref}$ $I_o = 1A$	50	56		dB
f	Switching frequency		85	100	115	KHz
$\frac{\Delta f}{\Delta V_i}$	Voltage stability of switching frequency	$V_i = 9V$ to 46V		0.5		%
$\frac{\Delta f}{\Delta T_j}$	Temperature stability of switching frequency	$T_j = 0\text{°C}$ to 125°C		1		%
f_{max}	Maximum operating switching frequency	$V_o = V_{ref}$ $I_o = 2A$	120	150		KHz
T_{sd}	Thermal shutdown junction temperature			150		°C

ELECTRICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
--------	-----------	-----------------	------	------	------	------

DC CHARACTERISTICS

I_{1Q}	Quiescent drain current	100% duty cycle pins 5 and 7 open	$V_i = 46V$		30	40	mA
		0% duty cycle			15	20	mA
$-I_{7L}$	Output leakage current	0% duty cycle				1	mA

SOFT START

I_{6SO}	Source current		100	140	180	μA
I_{6SI}	Sink current		50	70	120	μA

ERROR AMPLIFIER

V_{3H}	High level output voltage	$V_2 = 4.7V$	$I_3 = 100\mu A$	3.5			V
V_{3L}	Low level output voltage	$V_2 = 5.3V$	$I_3 = 100\mu A$			0.5	V
I_{3SI}	Sink output current	$V_2 = 5.3V$		100	150		μA
$-I_{3SO}$	Source output current	$V_2 = 4.7V$		100	150		μA
I_2	Input bias current	$V_2 = 5.2V$			2	10	μA
G_v	DC open loop gain	$V_3 = 1V$ to $3V$		46	55		dB

OSCILLATOR

$-I_5$	Oscillator source current		5			mA
--------	---------------------------	--	---	--	--	----

CIRCUIT OPERATION (refer to the block diagram)

The L4960 is a monolithic stepdown switching regulator providing output voltages from 5.1V to 40V and delivering 2.5A.

The regulation loop consists of a sawtooth oscillator, error amplifier, comparator and the output stage. An error signal is produced by comparing the output voltage with a precise 5.1V on-chip reference (zener zap trimmed to $\pm 2\%$).

This error signal is then compared with the sawtooth signal to generate the fixed frequency pulse width modulated pulses which drive the output stage.

The gain and frequency stability of the loop can be adjusted by an external RC network connected to pin 3. Closing the loop directly gives an output voltage of 5.1V. Higher voltages are obtained by inserting a voltage divider.

Output overcurrents at switch on are prevented by the soft start function. The error amplifier output is initially clamped by the external capacitor C_{SS} and

allowed to rise, linearly, as this capacitor is charged by a constant current source. Output overload protection is provided in the form of a current limiter. The load current is sensed by an internal metal resistor connected to a comparator. When the load current exceeds a preset threshold this comparator sets a flip flop which disables the output stage and discharges the soft start capacitor. A second comparator resets the flip flop when the voltage across the soft start capacitor has fallen to 0.4V.

The output stage is thus re-enabled and the output voltage rises under control of the soft start network. If the overload condition is still present the limiter will trigger again when the threshold current is reached. The average short circuit current is limited to a safe value by the dead time introduced by the soft start network. The thermal overload circuit disables circuit operation when the junction temperature reaches about 150°C and has hysteresis to prevent unstable conditions.

Figure 1. Soft start waveforms

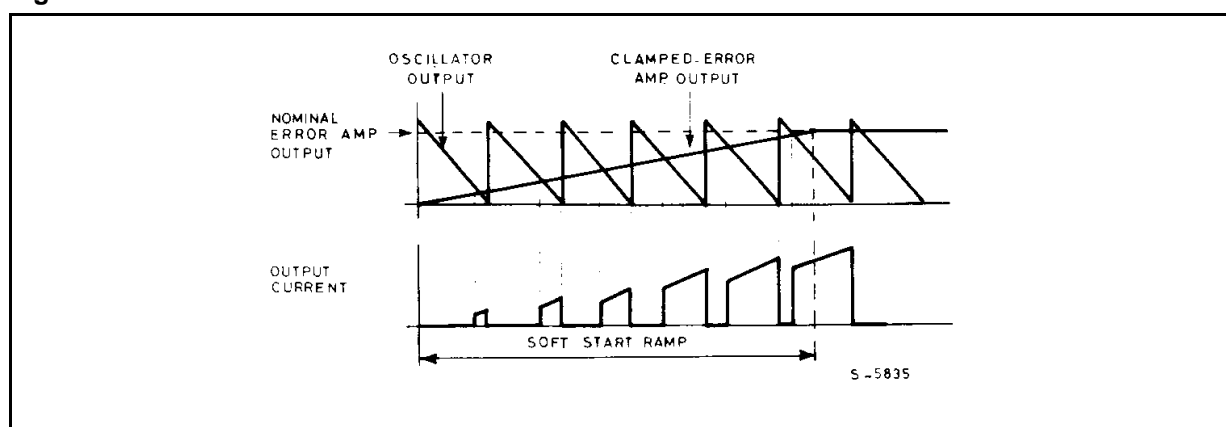
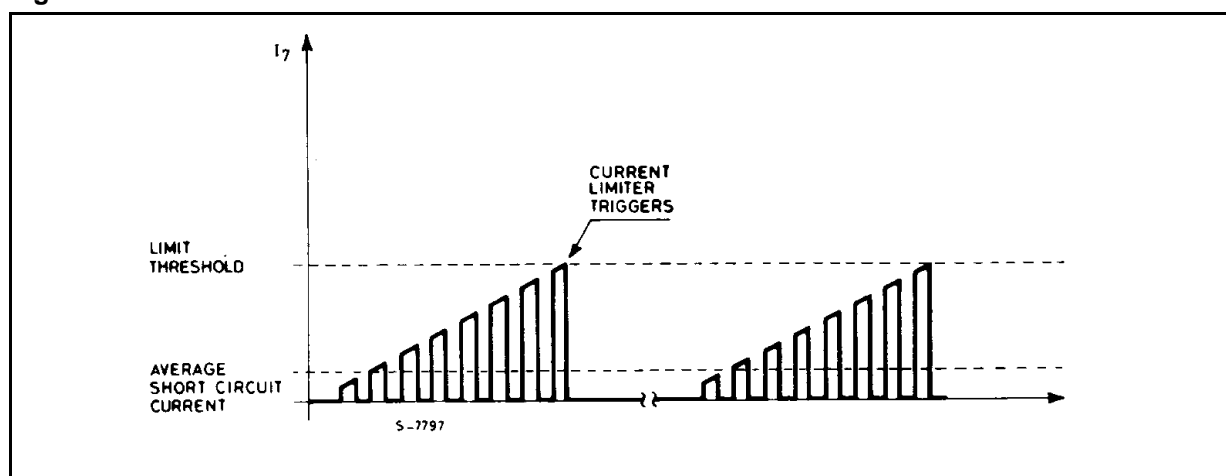
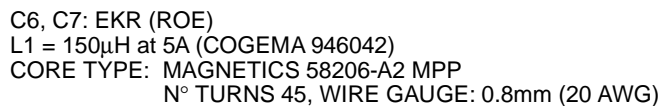


Figure 2. Current limiter waveforms





Graph of collector current I_Q (mA) versus collector voltage V_C (V) for the common-emitter circuit. The curve shows a slight increase in current with voltage, leveling off around 15 mA. The graph is labeled G-6065.

Graph of collector current I_C (mA) versus collector voltage V_C (V) for G-6066. The curve shows a slight increase in current with voltage, starting around 29 mA at 10 V and reaching about 31 mA at 40 V.

Figure 1 is a graph showing the temperature dependence of the current I_Q (mA) for a GaAs MESFET. The y-axis is labeled I_Q (mA) and ranges from 1.1 to 19. The x-axis is labeled T_j (°C) and ranges from -25 to 125. A curve shows I_Q increasing from approximately 14.5 mA at -25°C to about 16.5 mA at 125°C. Text in the graph indicates $G = 6067$ and $V_i = 35V$.

Figure 7. Quiescent drain current vs. junction temperature (100% duty cycle)

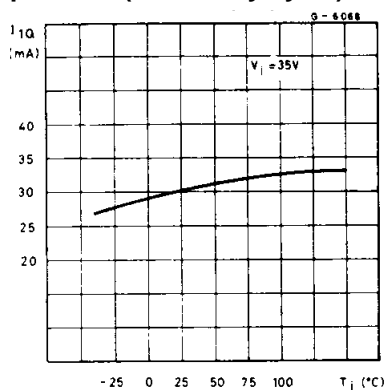


Figure 8. Reference voltage (pin 2) vs. V_i

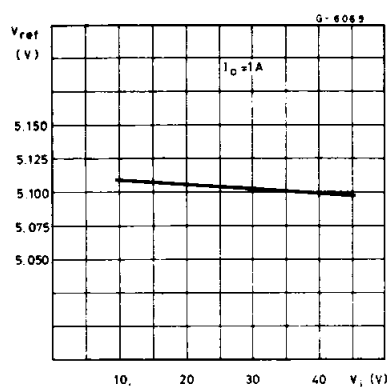


Figure 9. Reference voltage versus junction temperature (pin 2)

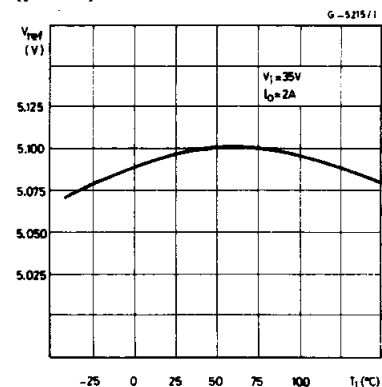


Figure 10. Open loop frequency and phase response of error amplifier

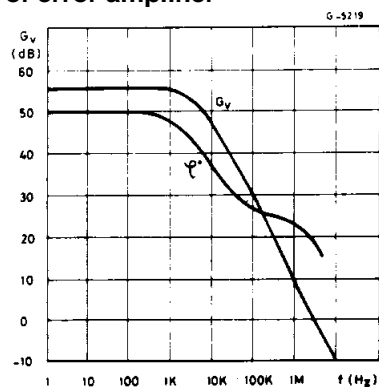


Figure 11. Switching frequency vs. input voltage

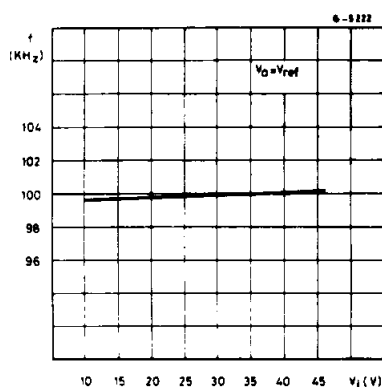


Figure 12. Switching frequency vs. junction temperature

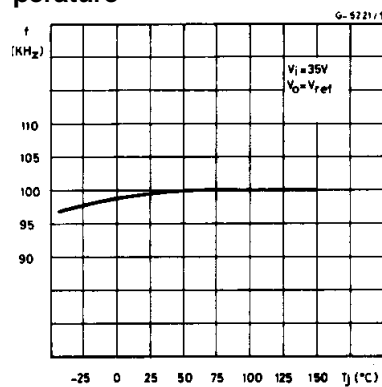


Figure 13. Switching frequency vs. R2 (see test circuit)

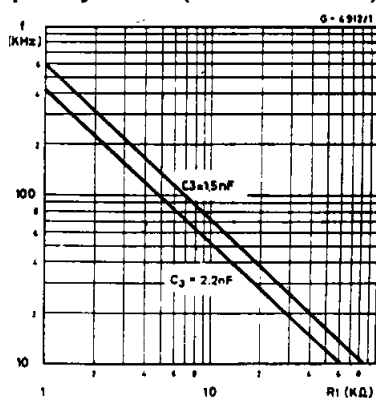


Figure 14. Line transient response

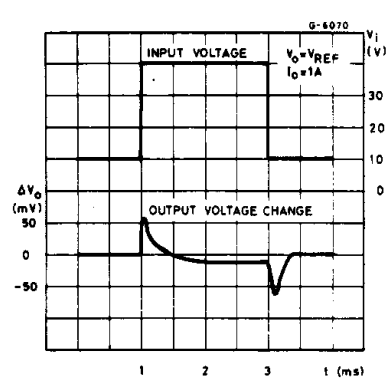


Figure 15. Load transient response

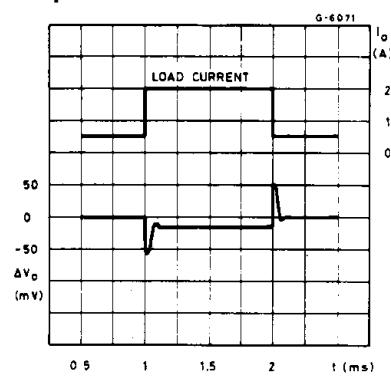


Figure 16. Supply voltage ripple rejection vs. frequency

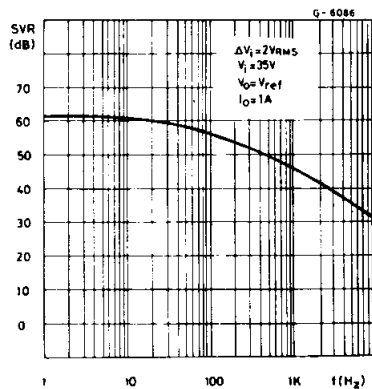


Figure 17. Dropout voltage between pin 1 and pin 7 vs. current at pin 7

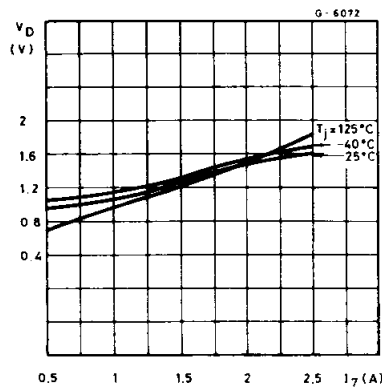


Figure 18. Dropout voltage between pin 1 and 7 vs. junction temperature

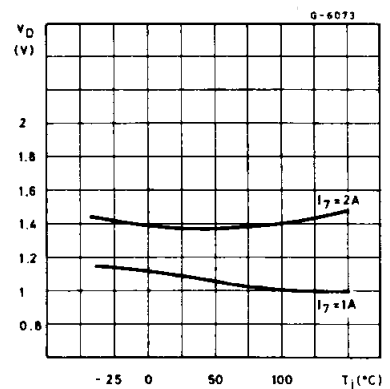


Figure 19. Power dissipation derating curve

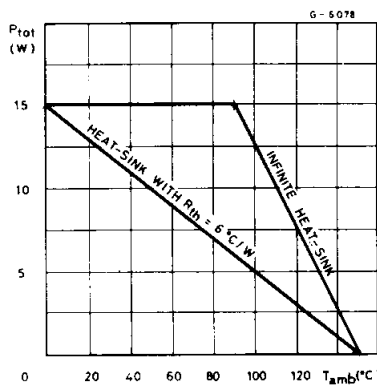


Figure 20. Efficiency vs. output current

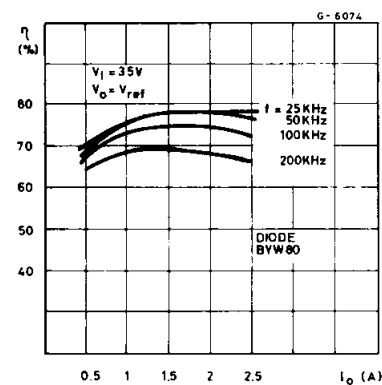


Figure 21. Efficiency vs. output current

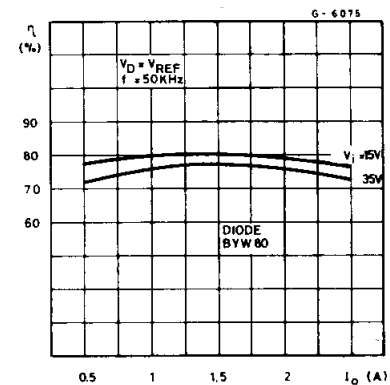


Figure 22. Efficiency vs. output current

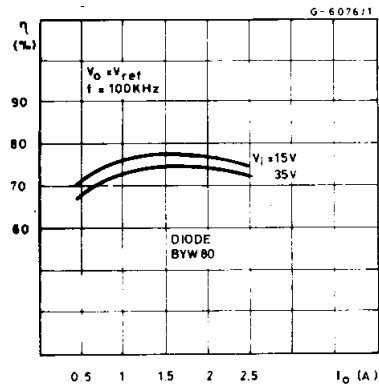
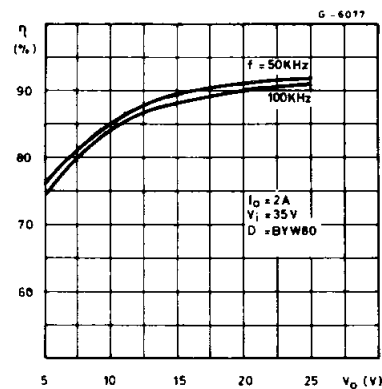


Figure 23. Efficiency vs. output voltage



APPLICATION INFORMATION

Figure 24. Typical application circuit

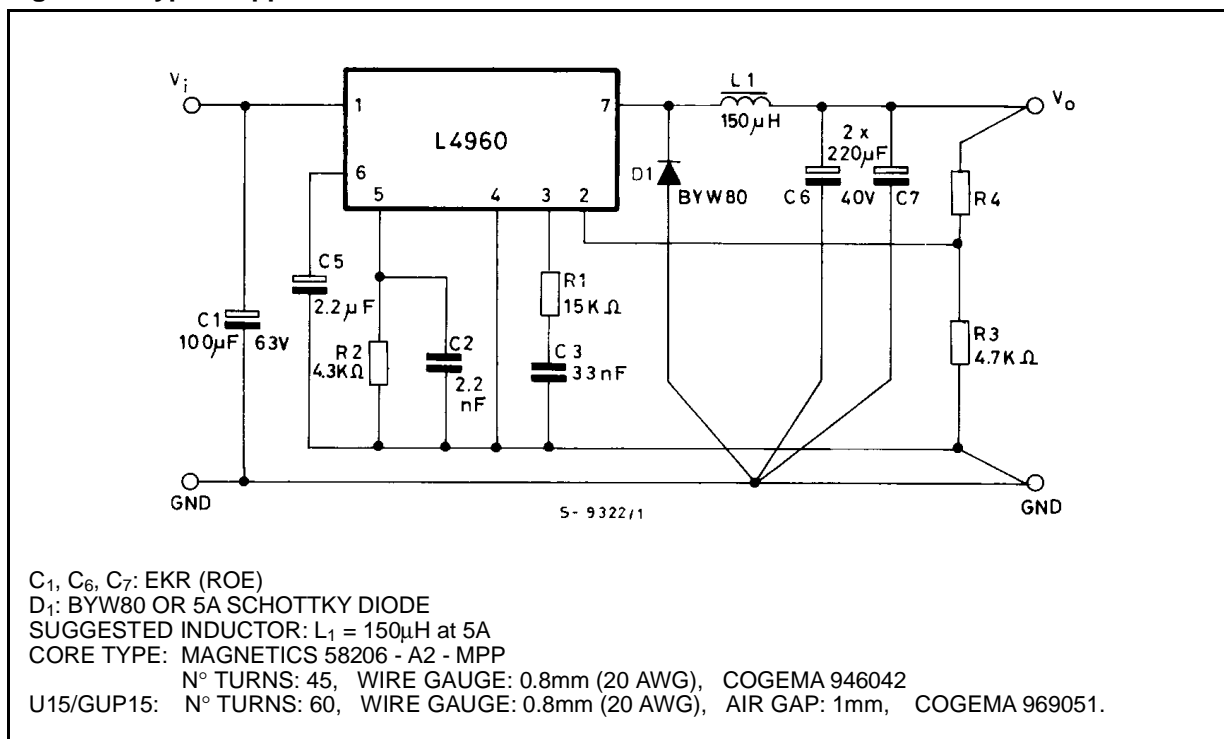
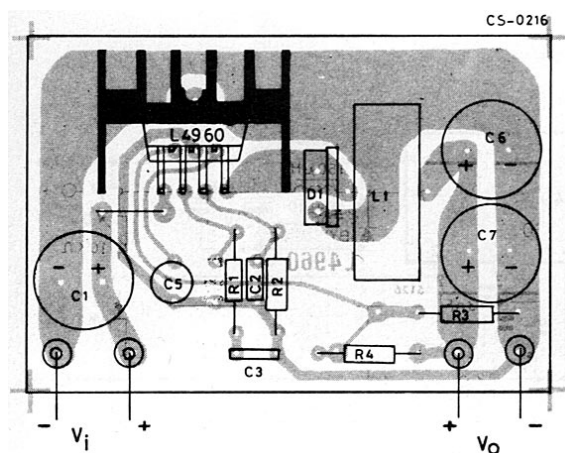


Figure 25. P.C. board and component layout of the Fig. 24 (1 : 1 scale)



Resistor values for standard output voltages		
V_o	R3	R4
12V	4.7KΩ	6.2KΩ
15V	4.7KΩ	9.1KΩ
18V	4.7KΩ	12KΩ
24V	4.7KΩ	18KΩ

APPLICATION INFORMATION

Figure 26. A minimal 5.1V fixed regulator; Very few component are required

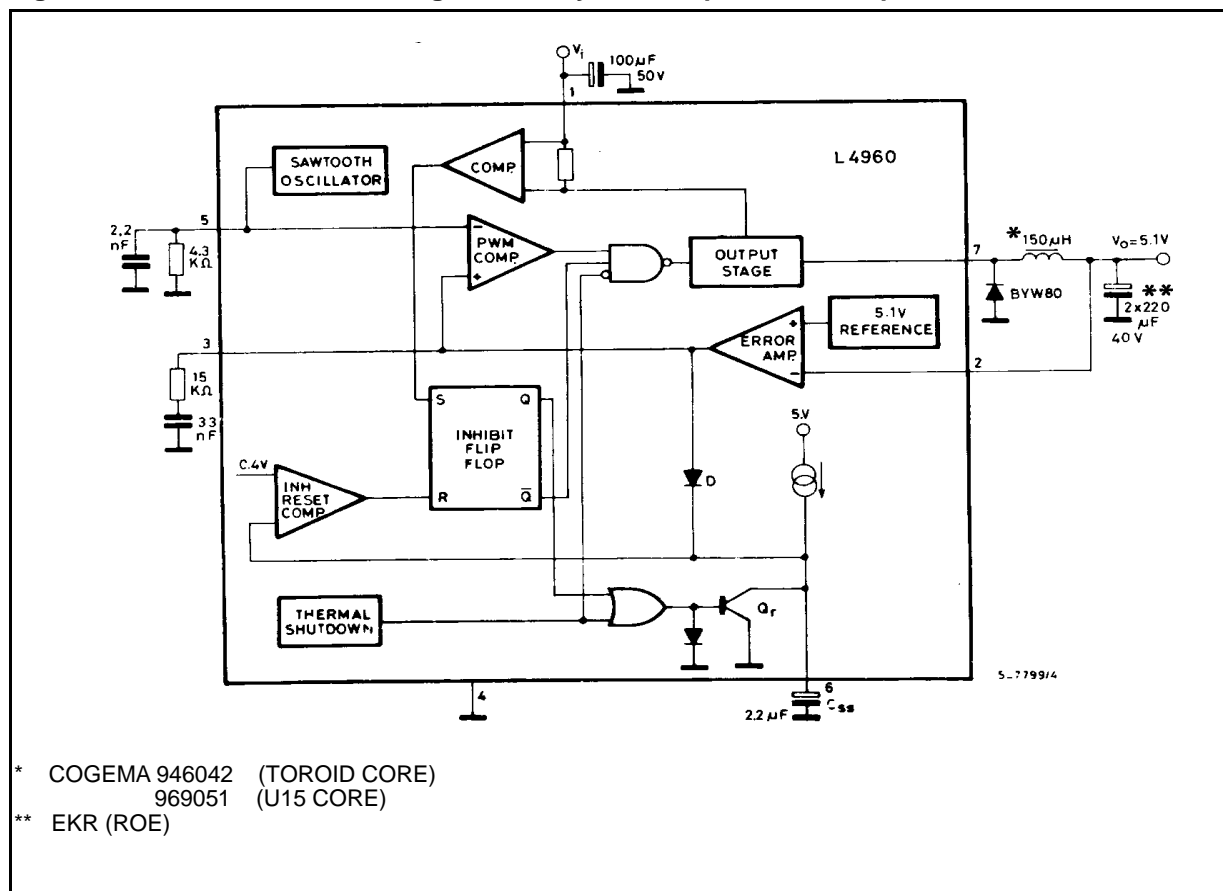
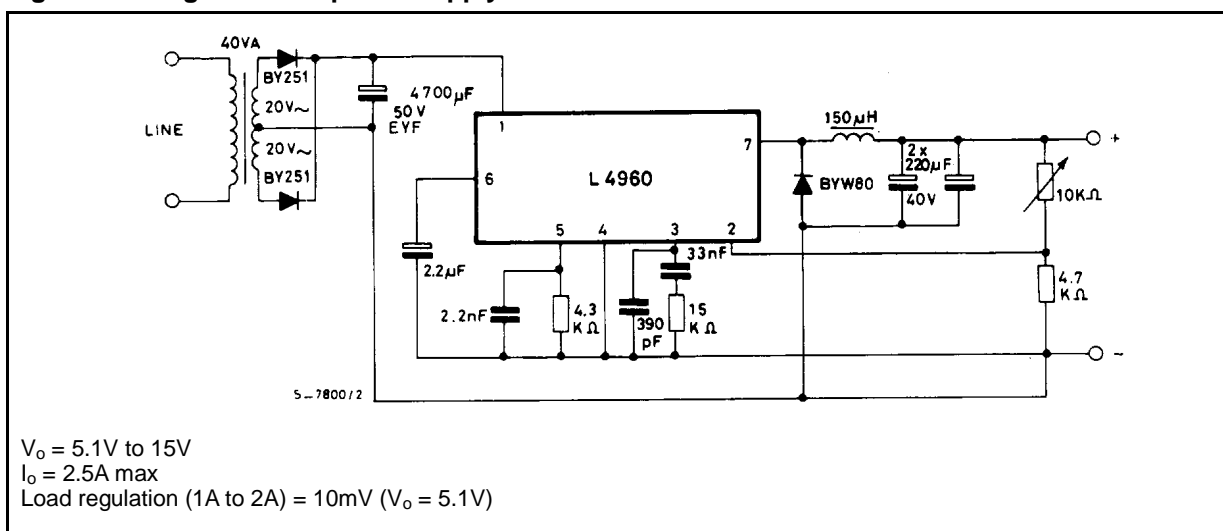
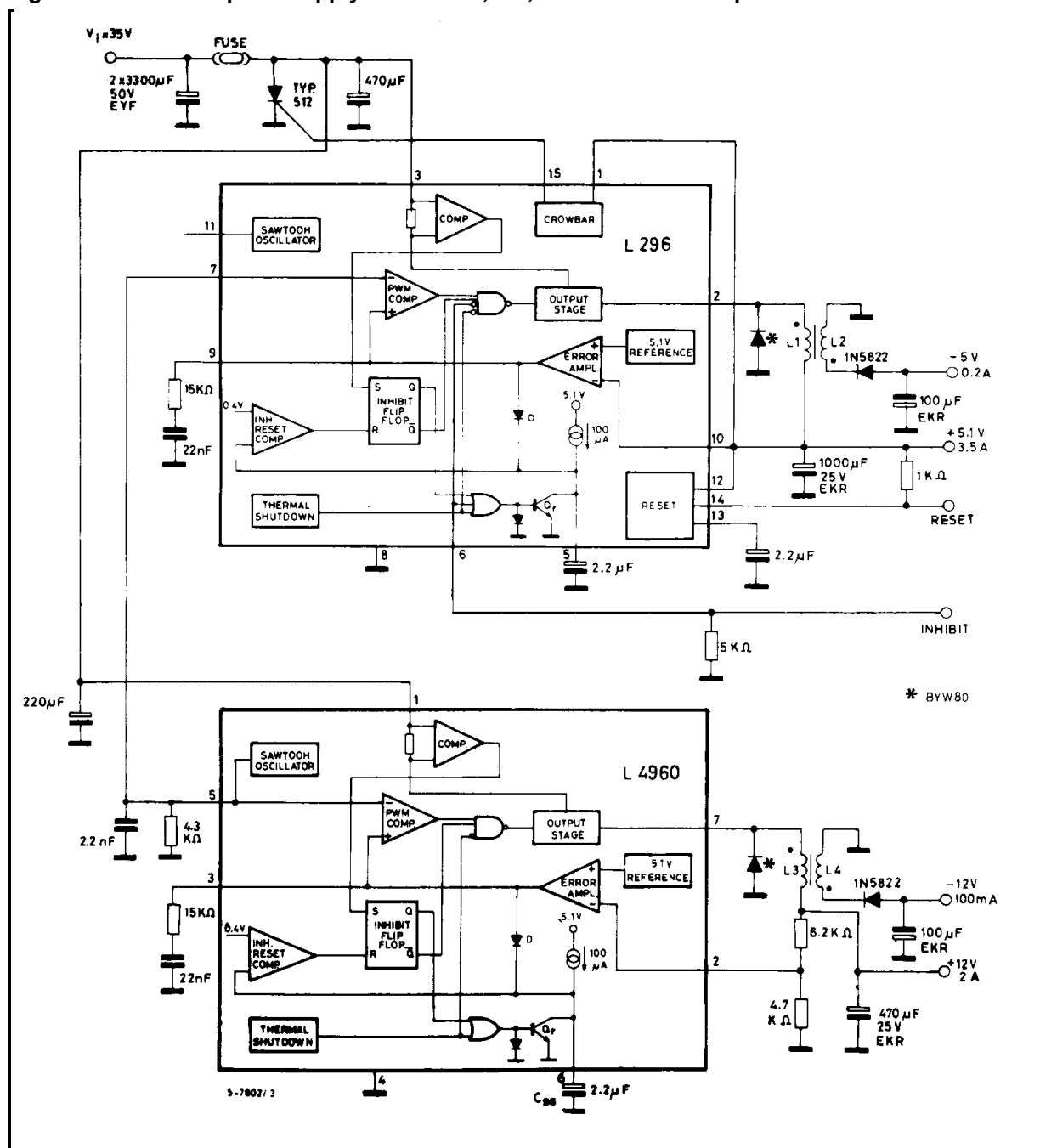


Figure 27. Programmable power supply



APPLICATION INFORMATION (continued)

Figure 28. Microcomputer supply with + 5.1V, -5V, +12V and -12V outputs



APPLICATION INFORMATION (continued)

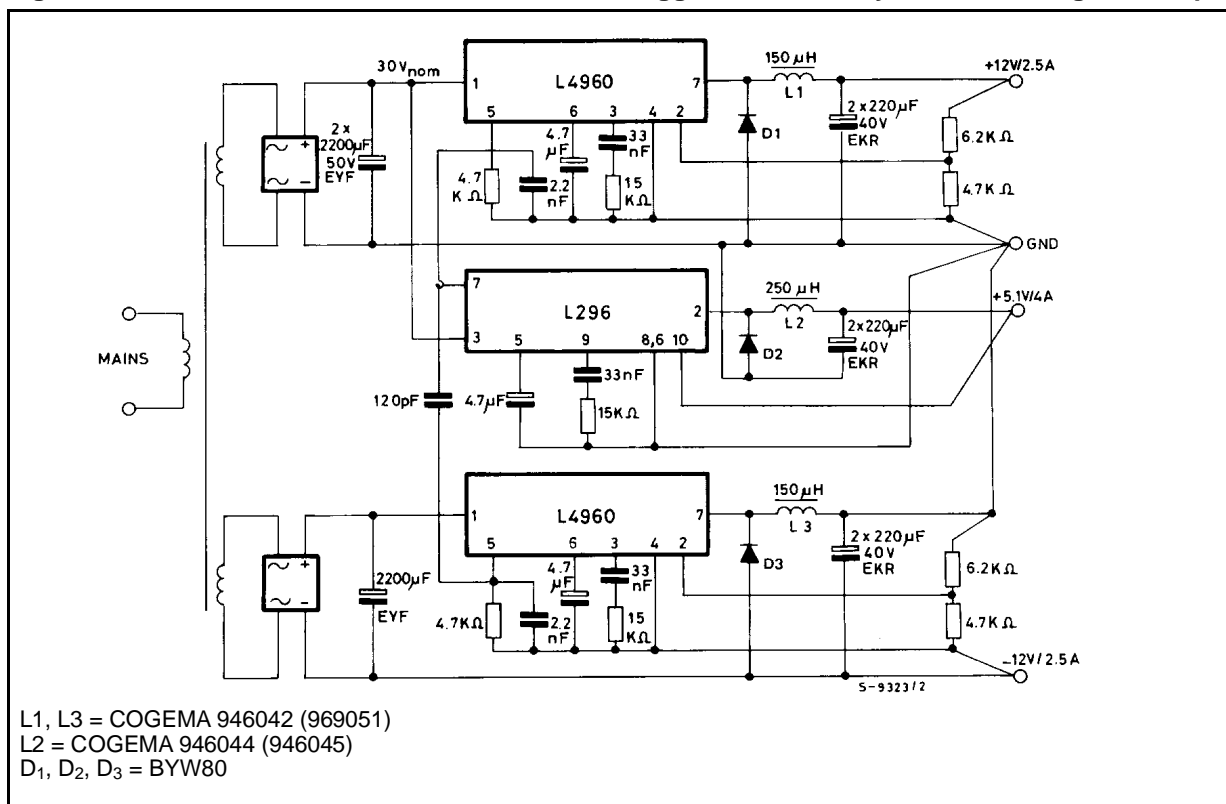
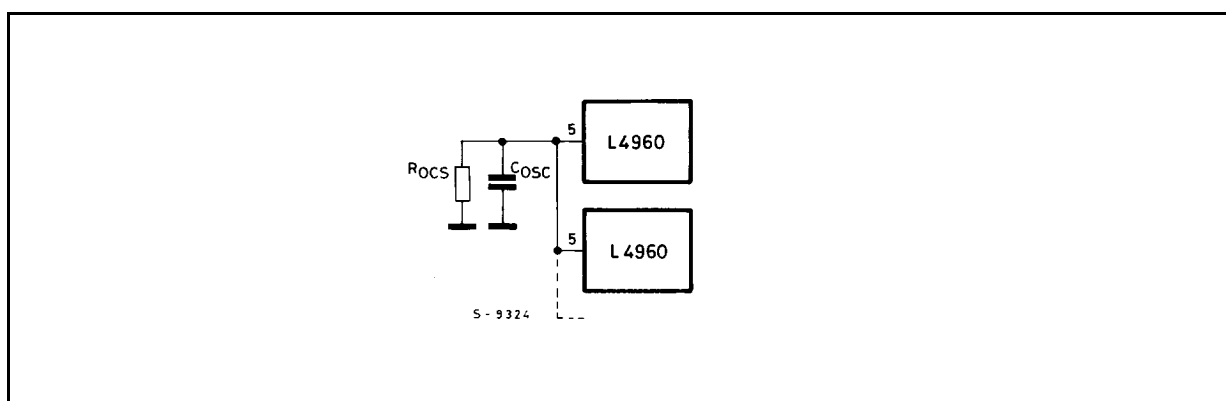
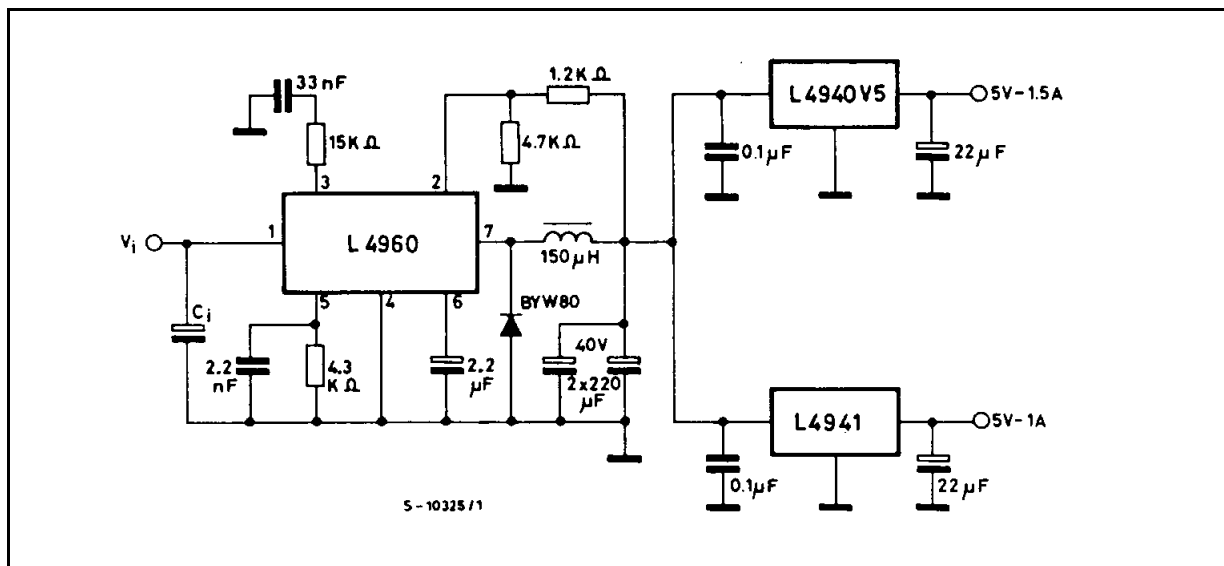
Figure 29. DC-DC converter 5.1V/4A, $\pm 12\text{V}/2.5\text{A}$; a suggestion how to synchronize a negative output

Figure 30. - In multiple supplies several L4960s can be synchronized as shown



APPLICATION INFORMATION (continued)

Figure 31. Regulator for distributed supplies

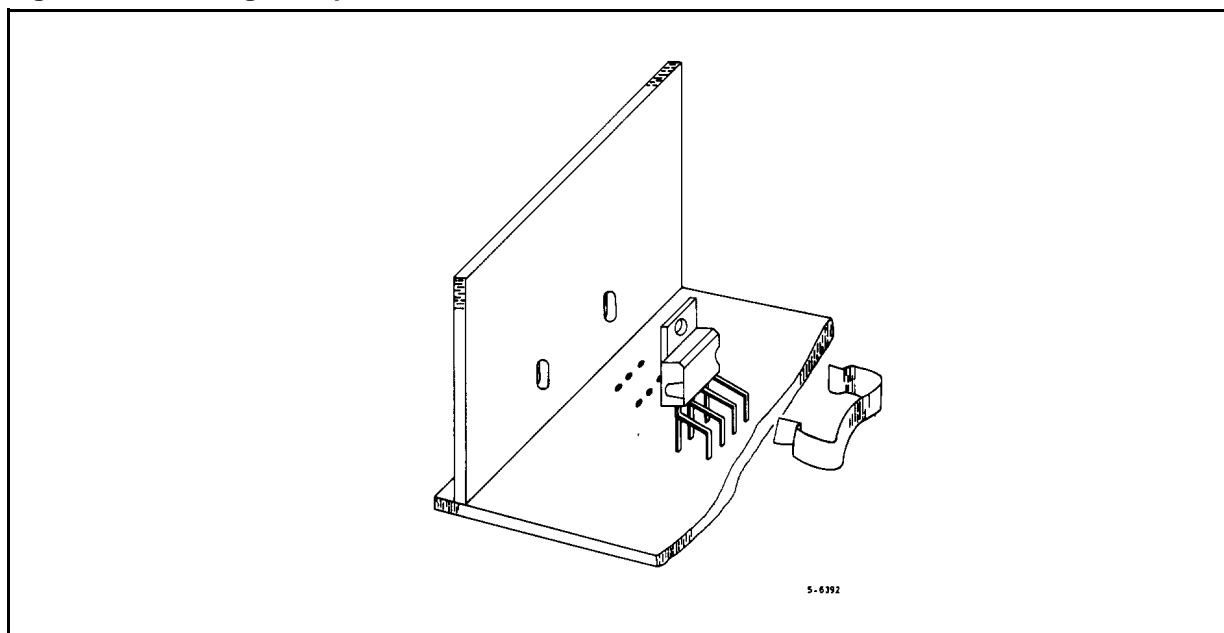


MOUNTING INSTRUCTION

The power dissipated in the circuit must be removed by adding an external heatsink. Thanks to the Heptawatt package attaching the heatsink is very simple, a screw or a compression spring (clip) being sufficient. Between the heatsink

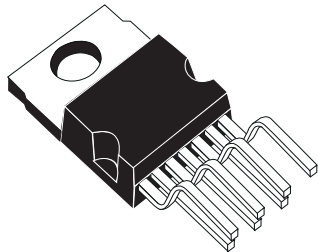
and the package it is better to insert a layer of silicon grease, to optimize the thermal contact, no electrical isolation is needed between the two surfaces.

Figure 32. Mounting example

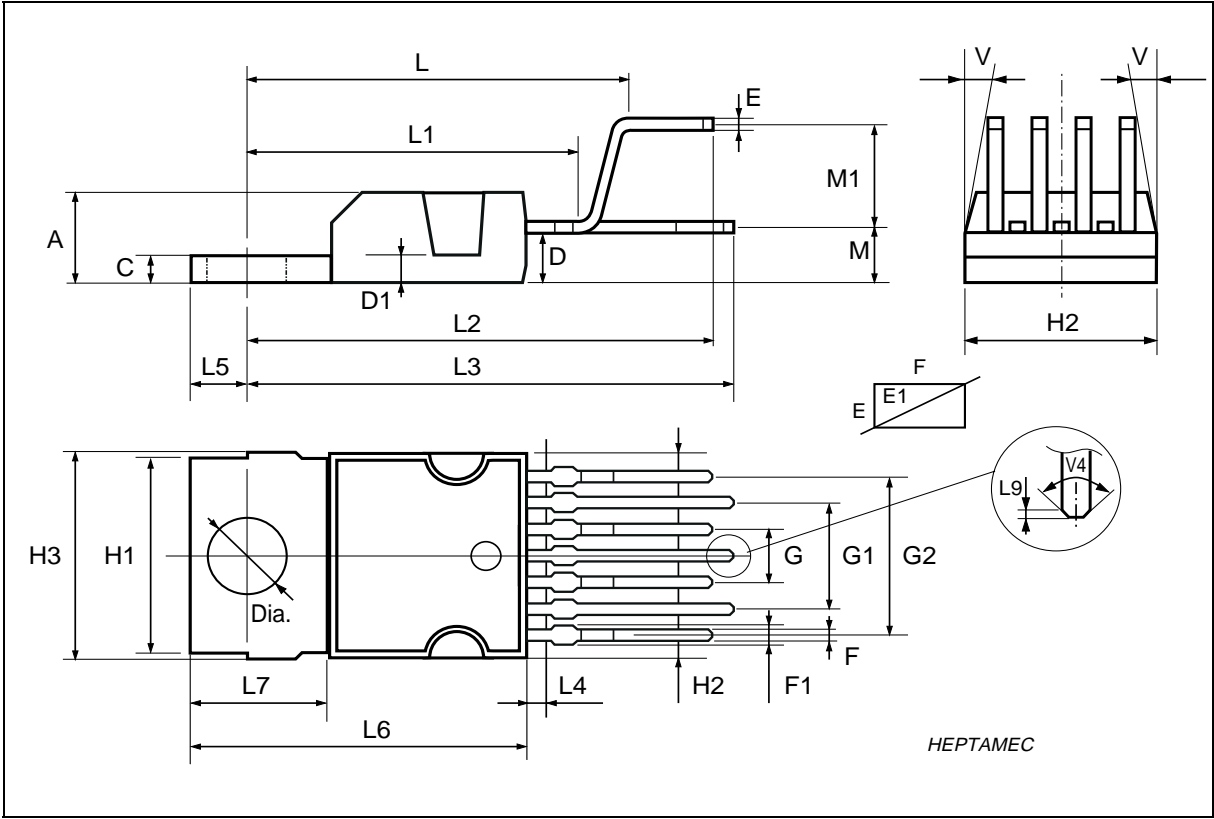


DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			4.8			0.189
C			1.37			0.054
D	2.4		2.8	0.094		0.110
D1	1.2		1.35	0.047		0.053
E	0.35		0.55	0.014		0.022
E1	0.7		0.97	0.028		0.038
F	0.6		0.8	0.024		0.031
F1			0.9			0.035
G	2.34	2.54	2.74	0.095	0.100	0.105
G1	4.88	5.08	5.28	0.193	0.200	0.205
G2	7.42	7.62	7.82	0.295	0.300	0.307
H2			10.4			0.409
H3	10.05		10.4	0.396		0.409
L	16.7	16.9	17.1	0.657	0.668	0.673
L1		14.92			0.587	
L2	21.24	21.54	21.84	0.386	0.848	0.860
L3	22.27	22.52	22.77	0.877	0.891	0.896
L4			1.29			0.051
L5	2.6	2.8	3	0.102	0.110	0.118
L6	15.1	15.5	15.8	0.594	0.610	0.622
L7	6	6.35	6.6	0.236	0.250	0.260
L9		0.2			0.008	
M	2.55	2.8	3.05	0.100	0.110	0.120
M1	4.83	5.08	5.33	0.190	0.200	0.210
V4	40° (typ.)					
Dia	3.65		3.85	0.144		0.152

**OUTLINE AND
MECHANICAL DATA**

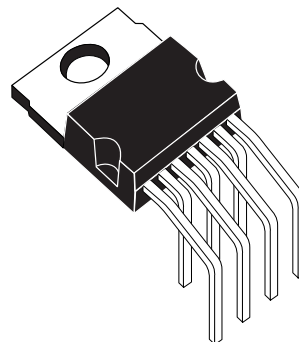


Heptawatt V

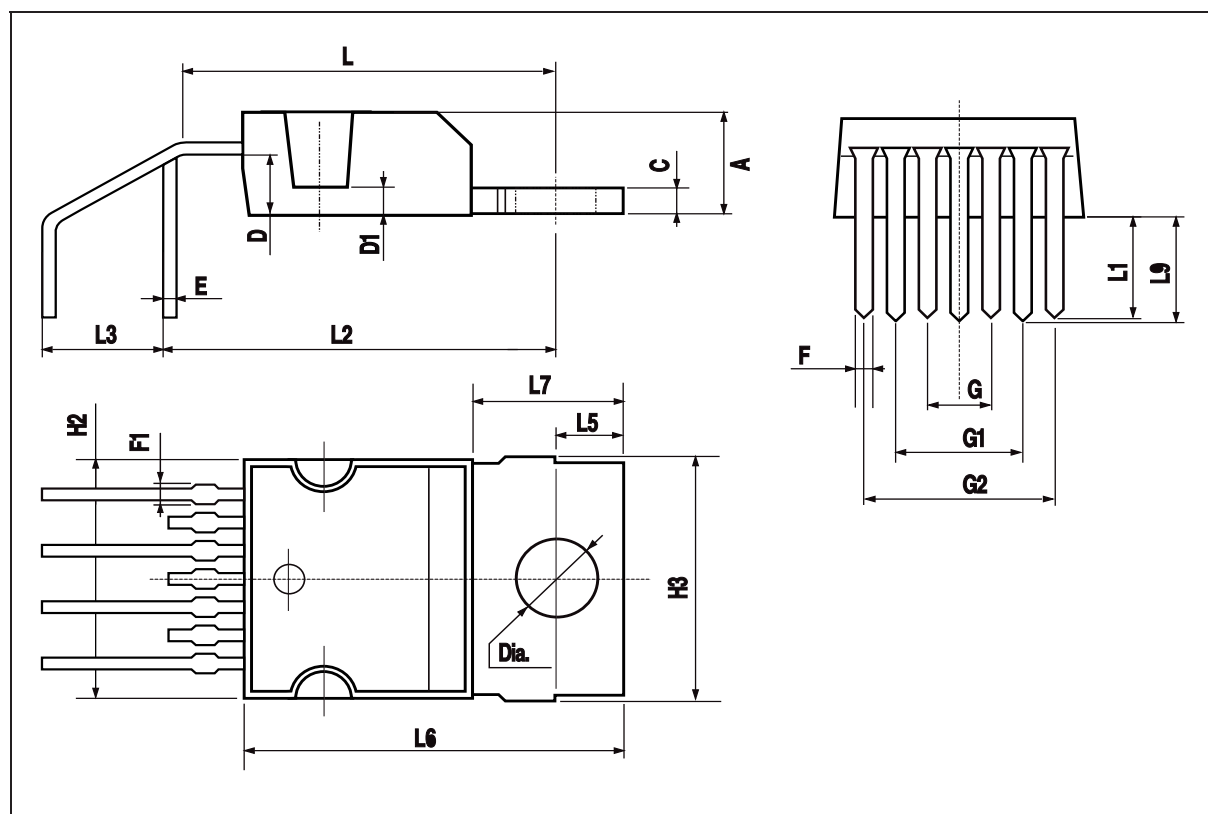


DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			4.8			0.189
C			1.37			0.054
D	2.4		2.8	0.094		0.110
D1	1.2		1.35	0.047		0.053
E	0.35		0.55	0.014		0.022
F	0.6		0.8	0.024		0.031
F1			0.9			0.035
G	2.41	2.54	2.67	0.095	0.100	0.105
G1	4.91	5.08	5.21	0.193	0.200	0.205
G2	7.49	7.62	7.8	0.295	0.300	0.307
H2			10.4			0.409
H3	10.05		10.4	0.396		0.409
L		14.2			0.559	
L1		4.4			0.173	
L2		15.8			0.622	
L3		5.1			0.201	
L5	2.6		3	0.102		0.118
L6	15.1		15.8	0.594		0.622
L7	6		6.6	0.236		0.260
L9		4.44			0.175	
Dia	3.65		3.85	0.144		0.152

OUTLINE AND MECHANICAL DATA



Heptawatt H



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