

## ISO154x Low-Power Bidirectional I<sup>2</sup>C Isolators

### 1 Features

- Isolated Bidirectional, I<sup>2</sup>C Compatible, Communication
- Supports up to 1-MHz Operation
- 3-V to 5.5-V Supply Range
- Open-Drain Outputs With 3.5-mA Side 1 and 35-mA Side 2 Sink Current Capability
- –40°C to +125°C Operating Temperature
- ±50-kV/μs Transient Immunity (Typical)
- HBM ESD Protection of 4 kV on All Pins; 8 kV on Bus Pins
- Safety-Related Certifications:
  - 4242-V<sub>PK</sub> Isolation per DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12
  - 2500-V<sub>RMS</sub> Isolation for 1 Minute per UL 1577
  - CSA Component Acceptance Notice 5A, IEC 60950-1 and IEC 61010-1 End Equipment Standards
  - CQC Basic Insulation per GB4943.1-2011

### 2 Applications

- Isolated I<sup>2</sup>C Buses
- SMBus and PMBus Interfaces
- Open-Drain Networks
- Motor Control Systems
- Battery Management
- I<sup>2</sup>C Level Shifting

### 3 Description

The ISO1540 and ISO1541 devices are low-power, bidirectional isolators that are compatible with I<sup>2</sup>C interfaces. These devices have logic input and output buffers that are separated by Texas Instruments Capacitive Isolation technology using a silicon dioxide (SiO<sub>2</sub>) barrier. When used with isolated power supplies, these devices block high voltages, isolate grounds, and prevent noise currents from entering the local ground and interfering with or damaging sensitive circuitry.

This isolation technology provides for function, performance, size, and power consumption advantages when compared to optocouplers. The ISO1540 and ISO1541 devices enable a complete isolated I<sup>2</sup>C interface to be implemented within a small form factor.

The ISO1540 has two isolated bidirectional channels for clock and data lines while the ISO1541 has a bidirectional data and a unidirectional clock channel. The ISO1541 is useful in applications that have a single master while the ISO1540 is suitable for multi-master applications. For applications where clock stretching by the slave is possible, the ISO1540 device should be used.

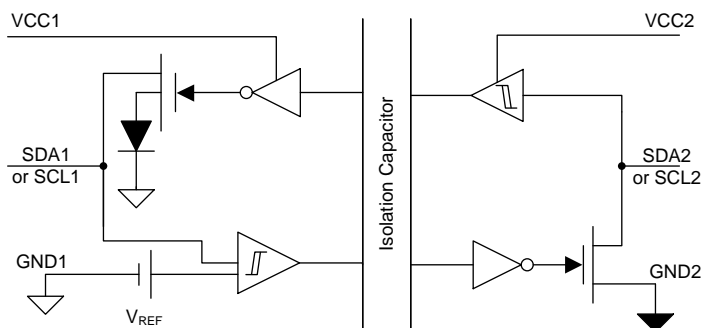
Isolated bidirectional communication is accomplished within these devices by offsetting the low-level output voltage on side 1 to a value greater than the high-level input voltage on side 1, thus preventing an internal logic latch that otherwise would occur with standard digital isolators.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
ISO1540 ISO1541	SOIC (8)	4.90 mm x 3.91 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Simplified Schematic

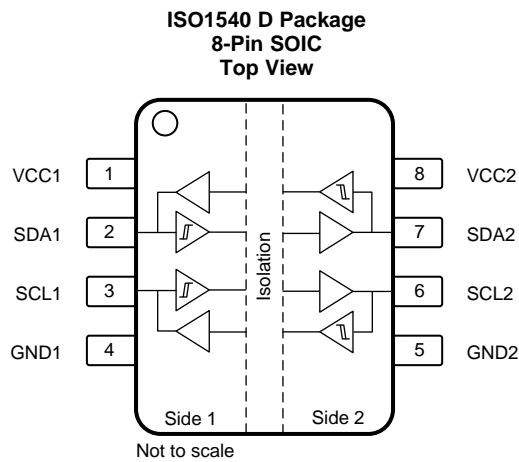




**Changes from Original (July 2012) to Revision A****Page**

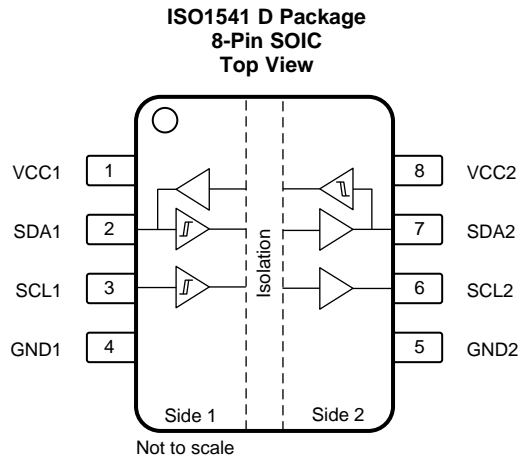
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  - Changed From: IEC 60950-1 and IEC 61010-1 End Equipment Standards (Pending) To: IEC 60950-1 and IEC 61010-1 End Equipment Standards (Approved)..... 1
  - Changed [Safety-Related Certifications](#), CSA column From: File number: 220991 (pending) To: File number: 220991..... 8
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## 5 Pin Configuration and Functions



**Pin Functions—ISO1540**

PIN		I/O	DESCRIPTION
NAME	NO.		
GND1	4	—	Ground, side 1
GND2	5	—	Ground, side 2
SCL1	3	I/O	Serial clock input / output, side 1
SCL2	6	I/O	Serial clock input / output, side 2
SDA1	2	I/O	Serial data input / output, side 1
SDA2	7	I/O	Serial data input / output, side 2
VCC1	1	—	Supply voltage, side 1
VCC2	8	—	Supply voltage, side 2



### Pin Functions—ISO1541

PIN		I/O	DESCRIPTION
NAME	NO.		
GND1	4	—	Ground, side 1
GND2	5	—	Ground, side 2
SCL1	3	I	Serial clock input, side 1
SCL2	6	O	Serial clock output, side 2
SDA1	2	I/O	Serial data input / output, side 1
SDA2	7	I/O	Serial data input / output, side 2
VCC1	1	—	Supply voltage, side 1
VCC2	8	—	Supply voltage, side 2

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)</sup>

		MIN	MAX	UNIT
Voltage	VCC1, VCC2	-0.5	6	V
	SDA1, SCL1	-0.5	VCC1 + 0.5 <sup>(3)</sup>	
	SDA2, SCL2	-0.5	VCC2 + 0.5 <sup>(3)</sup>	
I <sub>O</sub>	Output current			mA
		SDA1, SCL1	-20	
		SDA2, SCL2	-100	100
T <sub>J(MAX)</sub>	Maximum junction temperature		150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values here within are with respect to the local ground pin (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6 V.

## 6.2 ESD Ratings

			VALUE	UNIT	
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	Bus pins	±8000	V
			All pins	±4000	
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>		±1500	
		Machine Model JEDEC JESD22-A115-A, all pins		±200	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## 6.3 Recommended Operating Conditions

		MIN	MAX	UNIT
VCC1, VCC2	Supply voltage	3	5.5	V
V <sub>SDA1</sub> , V <sub>SCL1</sub>	Input and output signal voltages, side 1	0	VCC1	V
V <sub>SDA2</sub> , V <sub>SCL2</sub>	Input and output signal voltages, side 2	0	VCC2	V
V <sub>IL1</sub>	Low-level input voltage, side 1	0	0.5	V
V <sub>IH1</sub>	High-level input voltage, side 1	0.7 × VCC1	VCC1	V
V <sub>IL2</sub>	Low-level input voltage, side 2	0	0.3 × VCC2	V
V <sub>IH2</sub>	High-level input voltage, side 2	0.7 × VCC2	VCC2	V
I <sub>OL1</sub>	Output current, side 1	0.5	3.5	mA
I <sub>OL2</sub>	Output current, side 2	0.5	35	mA
C1	Capacitive load, side 1		40	pF
C2	Capacitive load, side 2		400	pF
f <sub>MAX</sub>	Operating frequency <sup>(1)</sup>		1	MHz
T <sub>A</sub>	Ambient temperature	−40	125	°C
T <sub>J</sub>	Junction temperature	−40	136	°C
T <sub>SD</sub>	Thermal shutdown	139	171	°C

(1) This represents the maximum frequency with the maximum bus load (C) and the maximum current sink (I<sub>O</sub>). If the system has less bus capacitance, then higher frequencies can be achieved.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		ISO154x	UNIT
		D (SOIC)	
		8 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	114.6	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	69.6	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	55.3	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	27.2	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	54.7	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report (SPRA953).

## 6.5 Power Ratings

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
P <sub>D</sub>	Maximum power dissipation (both sides)			85	mW	
P <sub>D1</sub>	Maximum power dissipation (side-1)	VCC1 = VCC2 = 5.5 V, T <sub>J</sub> = 150 °C, C1 = 40 pF, C2 = 400 pF;			34	mW
P <sub>D2</sub>	Maximum power dissipation (side-2)	Input a 1-MHz 50% duty cycle clock signal			51	mW

## 6.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
<b>GENERAL</b>				
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance through air	>4	mm
CPG	External creepage <sup>(1)</sup>	Shortest terminal-to-terminal distance across the package surface	>4	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	0.014	mm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>400	V
	Material group		II	
	Overvoltage category	Rated mains voltage $\leq 150 V_{RMS}$	I–IV	
		Rated mains voltage $\leq 300 V_{RMS}$	I–III	
<b>DIN V VDE V 0884-10 (VDE V 0884-10):2006-12<sup>(2)</sup></b>				
$V_{IORM}$	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	566	$V_{PK}$
$V_{IOTM}$	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$ $t = 60$ s (qualification) $t = 1$ s (100% production)	4242	$V_{PK}$
$q_{pd}$	Apparent charge <sup>(3)</sup>	Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60$ s; $V_{pd(m)} = 1.2 \times V_{IORM} = 680 V_{PK}$ , $t_m = 10$ s	<5	pC
		Method a: After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60$ s; $V_{pd(m)} = 1.6 \times V_{IORM} = 906 V_{PK}$ , $t_m = 10$ s	<5	
		Method b1: At routine test (100% production) and preconditioning (type test) $V_{ini} = V_{IOTM}$ , $t_{ini} = 1$ s; $V_{pd(m)} = 1.875 \times V_{IORM} = 1062 V_{PK}$ , $t_m = 1$ s	<5	
$C_{IO}$	Barrier capacitance, input to output <sup>(4)</sup>	$V_{IO} = 0.4 \sin(2\pi ft)$ , $f = 1$ MHz	~1	pF
$R_{IO}$	Isolation resistance, input to output <sup>(4)</sup>	$V_{IO} = 500$ V, $T_A = 25^\circ\text{C}$	$>10^{12}$	$\Omega$
		$V_{IO} = 500$ V, $100^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$	$>10^{11}$	
		$V_{IO} = 500$ V at $T_S = 150^\circ\text{C}$	$>10^9$	
	Pollution degree		2	
	Climatic category		40/125/21	
<b>UL 1577</b>				
$V_{ISO}$	Withstand isolation voltage	$V_{TEST} = V_{ISO} = 2500 V_{RMS}$ , $t = 60$ s (qualification); $V_{TEST} = 1.2 \times V_{ISO} = 3000 V_{RMS}$ , $t = 1$ s (100% production)	2500	$V_{RMS}$

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *basic electrical insulation* only within the maximum operating ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (4) All pins on each side of the barrier tied together creating a two-terminal device

## 6.7 Safety-Related Certifications

VDE	CSA	UL	CQC
Certified according to DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 and DIN EN 61010-1	Approved under CSA Component Acceptance Notice 5A, CSA/IEC 60950-1, and CSA/IEC 61010-1	Recognized under UL 1577 Component Recognition Program	Certified according to GB4943.1-2011
Basic Insulation Maximum Transient Overvoltage, 4242 V <sub>PK</sub> ; Maximum Repetitive Peak Voltage, 566 V <sub>PK</sub>	2.8-kV <sub>RMS</sub> Insulation Rating; 400 V <sub>RMS</sub> Basic Insulation working voltage per CSA 60950-1-07+A1+A2 and IEC 60950-1 2nd Ed.+A1+A2; 300 V <sub>RMS</sub> Basic, 150 V <sub>RMS</sub> Reinforced Insulation working voltage per CSA 61010-1-12 and IEC 61010-1 3rd Ed.,	Single protection, 2500 V <sub>RMS</sub>	Basic Insulation, Altitude ≤ 5000 m, Tropical Climate, 250 V <sub>RMS</sub> maximum working voltage
Certificate number: 40016131	Master contract number: 220991	File number: E181974	Certificate number: CQC14001109540

## 6.8 Safety Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier, potentially leading to secondary system failures.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>S</sub> Safety input, output, or supply current	R <sub>θJA</sub> = 114.6°C/W, V <sub>I</sub> = 5.5 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 1</a>			198	mA
	R <sub>θJA</sub> = 114.6°C/W, V <sub>I</sub> = 3.6 V, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C, see <a href="#">Figure 1</a>			303	
T <sub>S</sub> Safety temperature				150	°C

The safety-limiting constraint is the maximum junction temperature specified in the data sheet. The power dissipation and junction-to-air thermal impedance of the device installed in the application hardware determines the junction temperature. The assumed junction-to-air thermal resistance in the [Thermal Information](#) table is that of a device installed on a high-K test board for leaded surface-mount packages. The power is the recommended maximum input voltage times the current. The junction temperature is then the ambient temperature plus the power times the junction-to-air thermal resistance.



## 6.9 Electrical Characteristics

over recommended operating conditions, unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>SIDE 1 (ONLY)</b>						
$V_{ILT1}$	Voltage input threshold low, SDA1 and SCL1		500	550	660	mV
$V_{IHT1}$	Voltage input threshold high, SDA1 and SCL1		540	610	700	mV
$V_{HYST1}$	Voltage input hysteresis	$V_{IHT1} - V_{ILT1}$	40	60		mV
$V_{OL1}$	Low-level output voltage, SDA1 and SCL1 <sup>(1)</sup>	$0.5 \text{ mA} \leq (I_{SDA1} \text{ and } I_{SCL1}) \leq 3.5 \text{ mA}$	650		800	mV
$\Delta V_{OIT1}$	Low-level output voltage to high-level input voltage threshold difference, SDA1 and SCL1 <sup>(1)(2)</sup>	$0.5 \text{ mA} \leq (I_{SDA1} \text{ and } I_{SCL1}) \leq 3.5 \text{ mA}$	50			mV
<b>SIDE 2 (ONLY)</b>						
$V_{ILT2}$	Voltage input threshold low, SDA2 and SCL2		$0.3 \times V_{CC2}$	$0.4 \times V_{CC2}$		V
$V_{IHT2}$	Voltage input threshold high, SDA2 and SCL2		$0.4 \times V_{CC2}$	$0.5 \times V_{CC2}$		V
$V_{HYST2}$	Voltage input hysteresis	$V_{IHT2} - V_{ILT2}$	$0.05 \times V_{CC2}$			V
$V_{OL2}$	Low-level output voltage, SDA2 and SCL2	$0.5 \text{ mA} \leq (I_{SDA2} \text{ and } I_{SCL2}) \leq 35 \text{ mA}$			0.4	V
<b>BOTH SIDES</b>						
$ I_{l} $	Input leakage currents, SDA1, SCL1, SDA2, and SCL2	$V_{SDA1}, V_{SCL1} = V_{CC1};$ $V_{SDA2}, V_{SCL2} = V_{CC2}$		0.01	10	$\mu\text{A}$
$C_1$	Input capacitance to local ground, SDA1, SCL1, SDA2, and SCL2	$V_1 = 0.4 \times \sin(2E6\pi t) + 2.5 \text{ V}$		7		pF
CMTI	Common-mode transient immunity	See <a href="#">Figure 21</a>	25	50		kV/ $\mu\text{s}$
$V_{CCUV}$	VCC undervoltage lockout threshold <sup>(3)</sup>		2.1	2.5	2.8	V

(1) This parameter does not apply to the ISO1541 SCL1 line as it is unidirectional.

(2)  $\Delta V_{OIT1} = V_{OL1} - V_{IHT1}$ . This represents the minimum difference between a Low-Level Output Voltage and a High-Level Input Voltage Threshold to prevent a permanent latch condition that would otherwise exist with bidirectional communication.

(3) Any VCC voltages, on either side, less than the minimum will ensure device lockout. Both VCC voltages greater than the maximum will prevent device lockout.

## 6.10 Supply Current Characteristics

over recommended operating conditions, unless otherwise noted. For more information, see [Figure 19](#).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>3 V ≤ VCC1, VCC2 ≤ 3.6 V</b>						
I <sub>CC1</sub> Supply current, side 1	ISO1540	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		2.4	3.6	mA
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		2.5	3.8	
	ISO1541	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		2.1	3.3	
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		2.3	3.6	
I <sub>CC2</sub> Supply current, side 2	ISO1540 and ISO1541	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		1.7	2.7	mA
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		1.9	3.1	
<b>4.5 V ≤ VCC1, VCC2 ≤ 5.5 V</b>						
I <sub>CC1</sub> Supply current, side 1	ISO1540	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		3.1	4.7	mA
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		3.1	4.7	
	ISO1541	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		2.8	4.4	
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		2.9	4.5	
I <sub>CC2</sub> Supply current, side 2	ISO1540 and ISO1541	V <sub>SDA1</sub> , V <sub>SCL1</sub> = GND1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = GND2; R1, R2 = Open; C1, C2 = Open		2.3	3.7	mA
		V <sub>SDA1</sub> , V <sub>SCL1</sub> = VCC1; V <sub>SDA2</sub> , V <sub>SCL2</sub> = VCC2; R1, R2 = Open; C1, C2 = Open		2.5	4	

## 6.11 Timing Requirements

		MIN	NOM	MAX	UNIT	
t <sub>SP</sub>	Input noise filter	5	12		ns	
t <sub>UVLO</sub>	Time to recover from UVLO	2.7 V to 0.9 V; See <a href="#">Figure 22</a>		30	110	μs

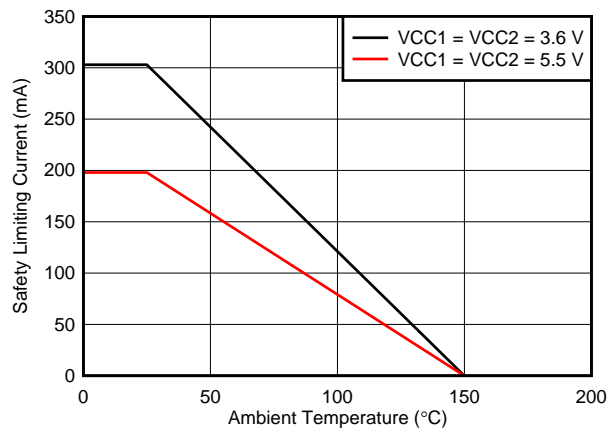
## 6.12 Switching Characteristics

over recommended operating conditions, unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT		
<b>3 V ≤ VCC1, VCC2 ≤ 3.6 V</b>								
t <sub>f1</sub>	Output Signal Fall Time (SDA1, SCL1)	See <a href="#">Figure 19</a> R1 = 953 Ω, C1 = 40 pF	0.7 × VCC1 to 0.3 × VCC1		8	17	29	ns
			0.9 × VCC1 to 900 mV		16	29	48	
t <sub>f2</sub>	Output Signal Fall Time (SDA2, SCL2)	See <a href="#">Figure 19</a> R2 = 95.3 Ω, C2 = 400 pF	0.7 × VCC2 to 0.3 × VCC2		14	23	47	ns
			0.9 × VCC2 to 400 mV		35	50	100	
t <sub>pLH1-2</sub>	Low-to-High Propagation Delay, Side 1 to Side 2	See <a href="#">Figure 19</a> R1 = 953 Ω, R2 = 95.3 Ω, C1, C2 = 10 pF	0.55 V to 0.7 × VCC2		33	65	ns	
t <sub>pHL1-2</sub>	High-to-Low Propagation Delay, Side 1 to Side 2		0.7 V to 0.4 V		90	181	ns	
PWD <sub>1-2</sub>	Pulse Width Distortion  t <sub>pHL1-2</sub> – t <sub>pLH1-2</sub>				55	123	ns	
t <sub>pLH2-1</sub> <sup>(1)</sup>	Low-to-High Propagation Delay, Side 2 to Side 1		0.4 × VCC2 to 0.7 × VCC1		47	68	ns	
t <sub>pHL2-1</sub> <sup>(1)</sup>	High-to-Low Propagation Delay, Side 2 to Side 1		0.4 × VCC2 to 0.9 V		67	109	ns	
PWD <sub>2-1</sub> <sup>(1)</sup>	Pulse Width Distortion  t <sub>pHL2-1</sub> – t <sub>pLH2-1</sub>				20	49	ns	
t <sub>LOOP1</sub> <sup>(1)</sup>	Round-trip propagation delay on Side 1		See <a href="#">Figure 20</a> ; R1 = 953 Ω, C1 = 40 pF R2 = 95.3 Ω, C2 = 400 pF	0.4 V to 0.3 × VCC1		100	165	ns
<b>4.5 V ≤ VCC1, VCC2 ≤ 5.5 V</b>								
t <sub>f1</sub>	Output Signal Fall Time (SDA1, SCL1)	See <a href="#">Figure 19</a> R1 = 1430 Ω, C1 = 40 pF	0.7 × VCC1 to 0.3 × VCC1		6	11	20	ns
			0.9 × VCC1 to 900 mV		13	21	39	
t <sub>f2</sub>	Output Signal Fall Time (SDA2, SCL2)	See <a href="#">Figure 19</a> R2 = 143 Ω, C2 = 400 pF	0.7 × VCC2 to 0.3 × VCC2		10	18	35	ns
			0.9 × VCC2 to 400 mV		28	41	76	
t <sub>pLH1-2</sub>	Low-to-High Propagation Delay, Side 1 to Side 2	See <a href="#">Figure 19</a> R1 = 1430 Ω, R2 = 143 Ω, C1,2 = 10 pF	0.55 V to 0.7 × VCC2		31	62	ns	
t <sub>pHL1-2</sub>	High-to-Low Propagation Delay, Side 1 to Side 2		0.7 V to 0.4 V		70	139	ns	
PWD <sub>1-2</sub>	Pulse Width Distortion  t <sub>pHL1-2</sub> – t <sub>pLH1-2</sub>				38	80	ns	
t <sub>pLH2-1</sub> <sup>(1)</sup>	Low-to-high propagation delay, side 2 to side 1		0.4 × VCC2 to 0.7 × VCC1		55	80	ns	
t <sub>pHL2-1</sub> <sup>(1)</sup>	High-to-low propagation delay, Side 2 to side 1		0.4 × VCC2 to 0.9 V		47	85	ns	
PWD <sub>2-1</sub> <sup>(1)</sup>	Pulse Width Distortion  t <sub>pHL2-1</sub> – t <sub>pLH2-1</sub>				8	21	ns	
t <sub>LOOP1</sub> <sup>(1)</sup>	Round-trip propagation delay on side 1		See <a href="#">Figure 20</a> ; R1 = 1430 Ω, C1 = 40 pF R2 = 143 Ω, C2 = 400 pF	0.4 V to 0.3 × VCC1		110	180	ns

(1) This parameter does not apply to the ISO1541 SCL1 line as it is unidirectional.

**6.13 Insulation Characteristics Curves**



**Figure 1. Thermal Derating Curve for Limiting Current per VDE**

## 6.14 Typical Characteristics

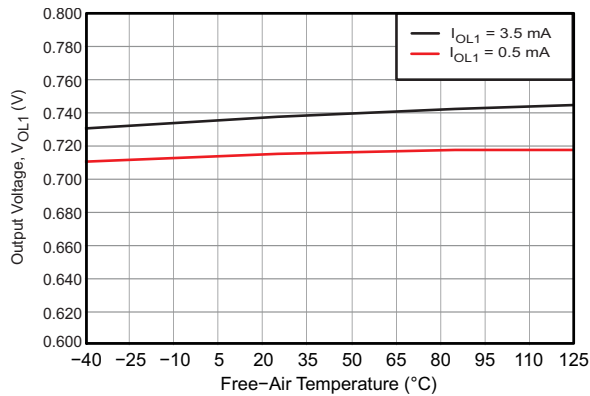


Figure 2. Side 1: Output Low Voltage vs Free-Air Temperature

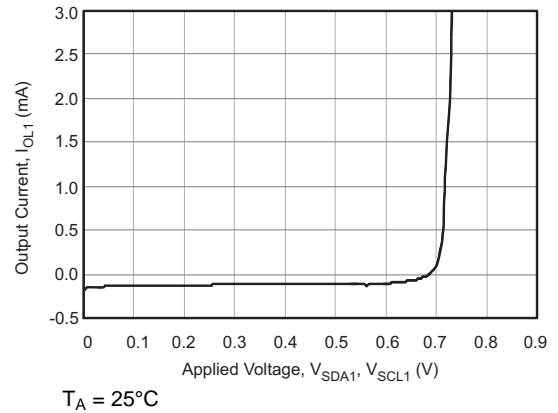


Figure 3. Side 1: Output Low Current vs  $S_{DA1}$  or  $S_{CL1}$  Applied Voltage

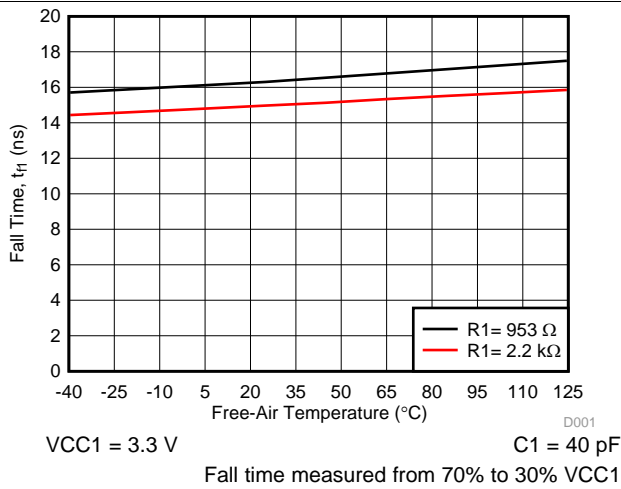


Figure 4. Side 1: Output Fall Time vs Free-Air Temperature

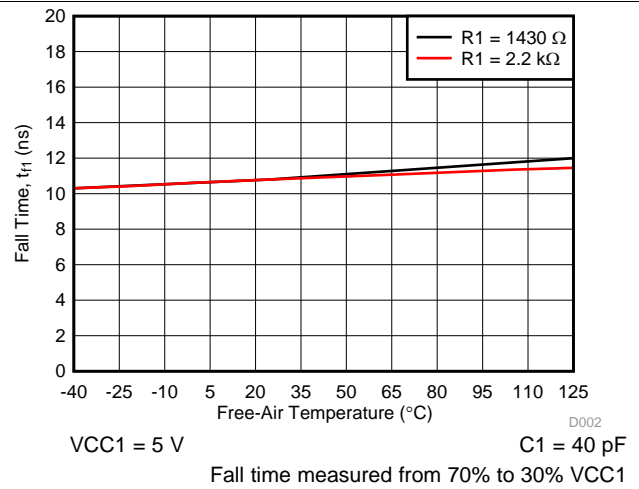


Figure 5. Side 1: Output Fall Time vs Free-air Temperature

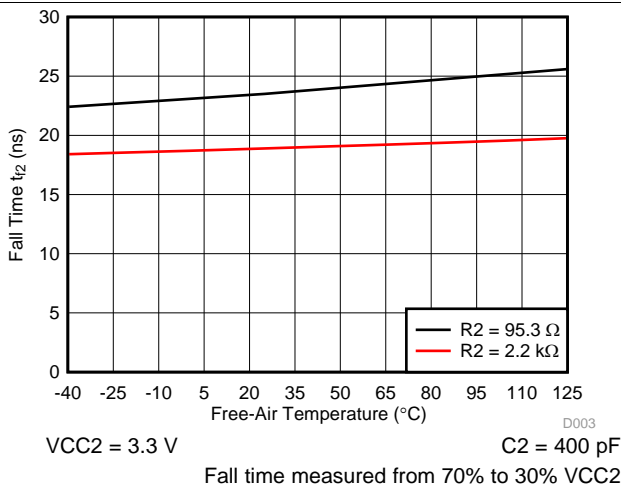


Figure 6. Side 2: Output Fall Time vs Free-Air Temperature

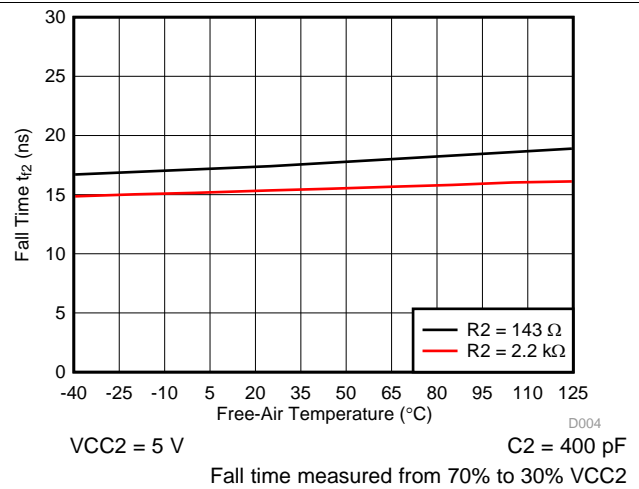


Figure 7. Side 2: Output Fall Time vs Free-Air Temperature

Typical Characteristics (continued)

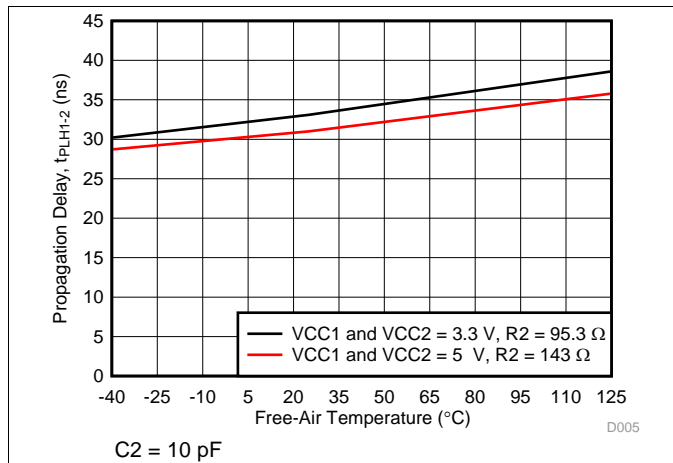


Figure 8.  $t_{PLH1-2}$  Propagation Delay vs Free-Air Temperature

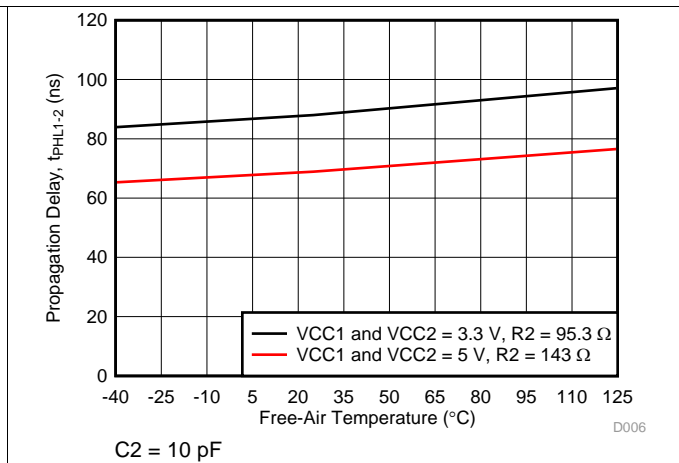


Figure 9.  $t_{PHL1-2}$  Propagation Delay vs Free-Air Temperature

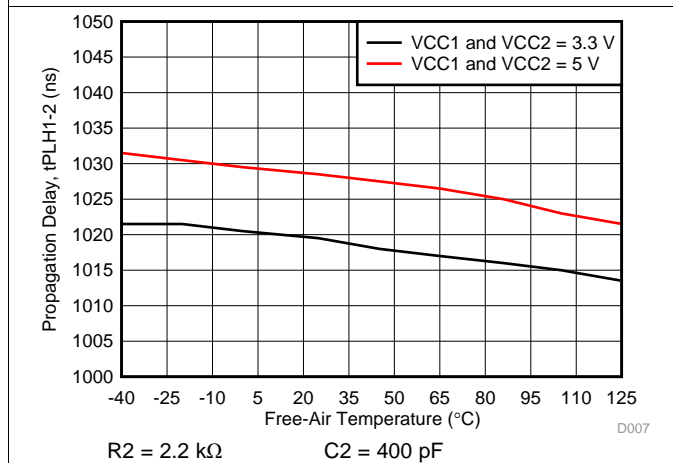


Figure 10.  $t_{PLH1-2}$  Propagation Delay vs Free-Air Temperature

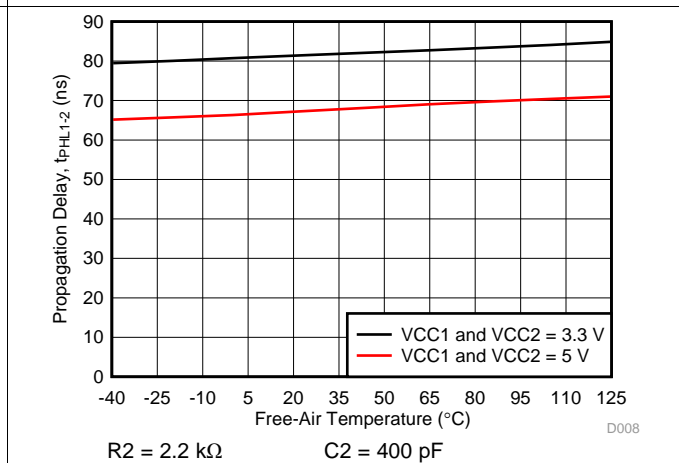


Figure 11.  $t_{PHL1-2}$  Propagation Delay vs Free-Air Temperature

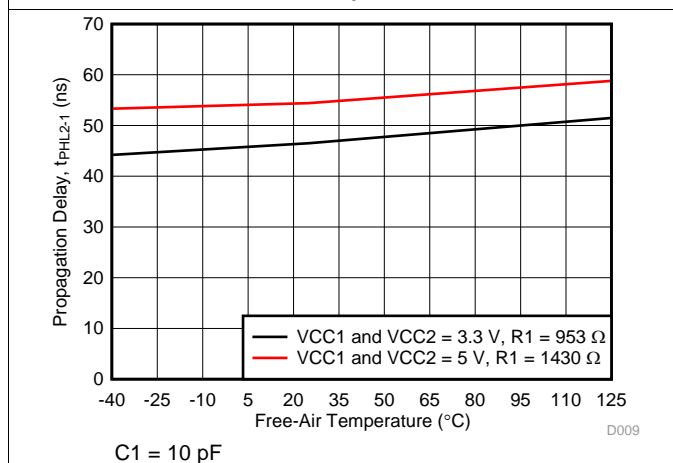


Figure 12.  $t_{PLH2-1}$  Propagation Delay vs Free-Air Temperature

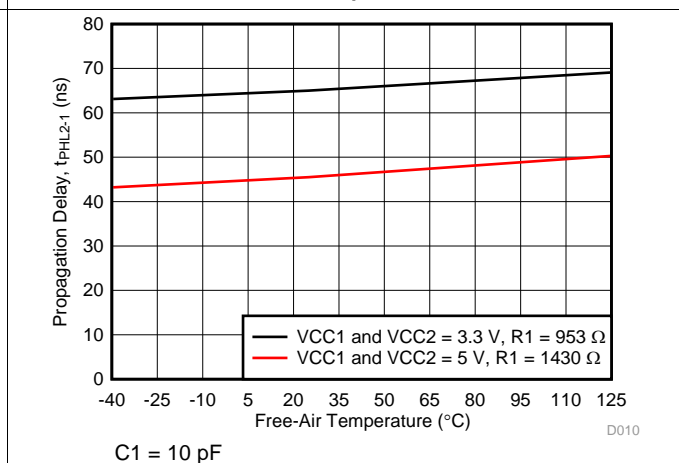


Figure 13.  $t_{PHL2-1}$  Propagation Delay vs Free-Air Temperature

Typical Characteristics (continued)

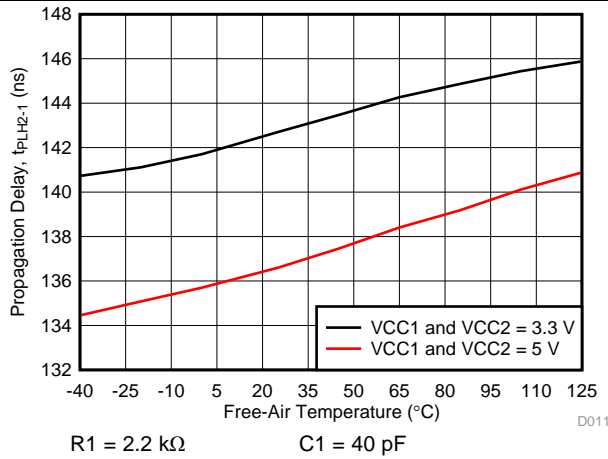


Figure 14.  $t_{PLH2-1}$  Propagation Delay vs Free-Air Temperature

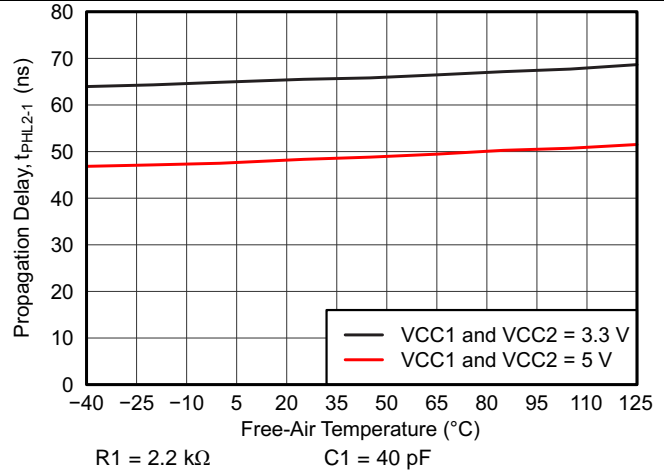


Figure 15.  $t_{PHL2-1}$  Propagation Delay vs Free-Air Temperature

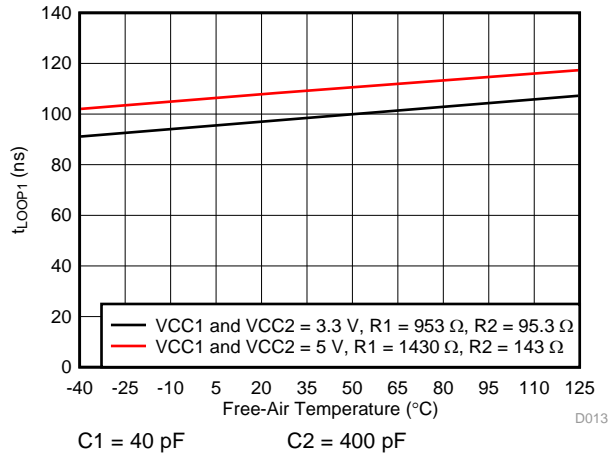


Figure 16.  $t_{LOOP1}$  vs Free-Air Temperature

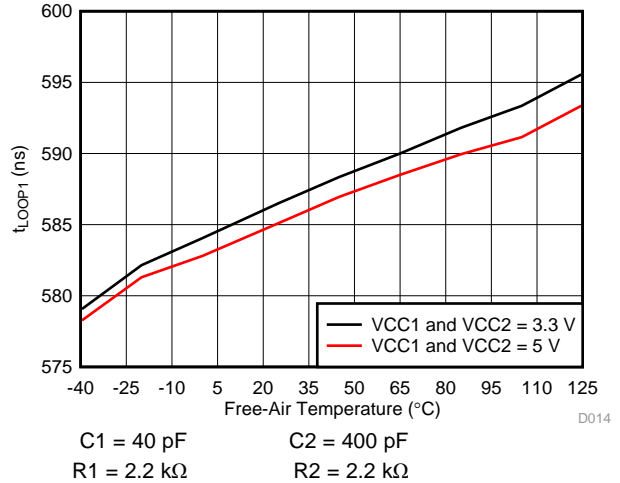


Figure 17.  $t_{LOOP1}$  vs Free-Air Temperature

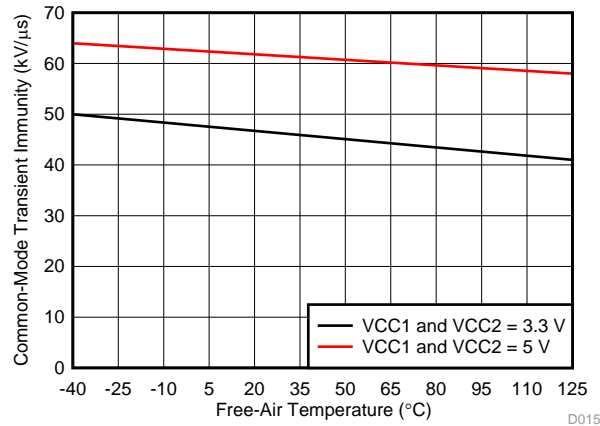
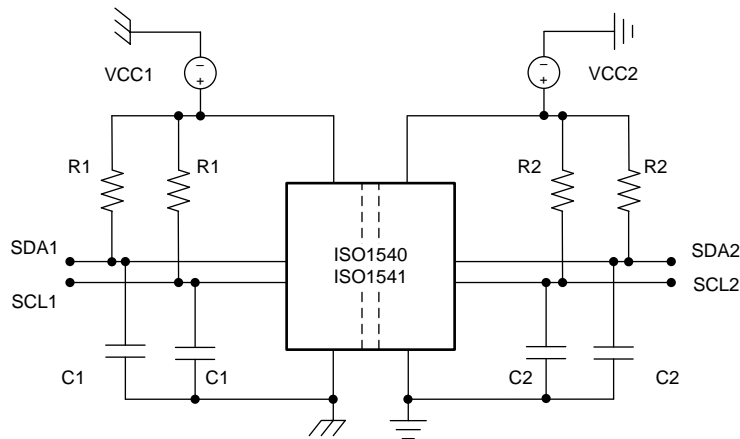


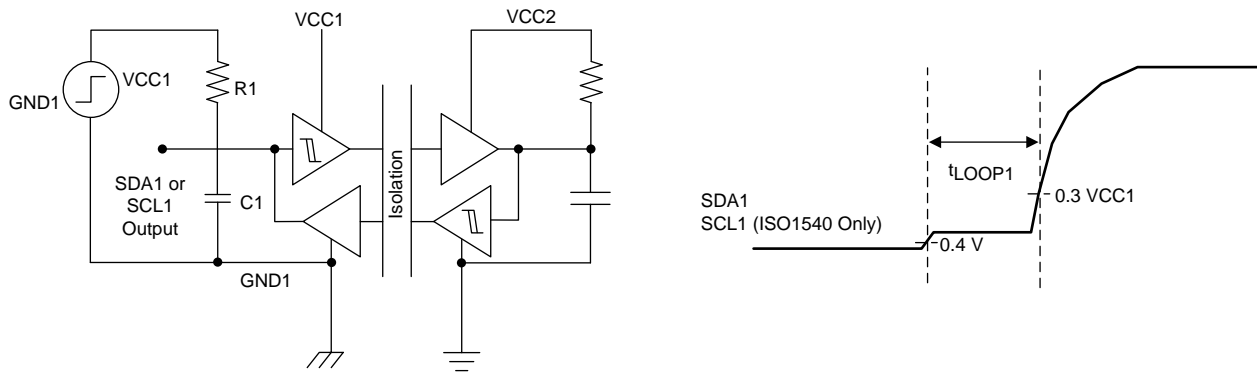
Figure 18. CMTI vs Free-Air Temperature

## 7 Parameter Measurement Information



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Figure 19. Test Diagram



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Figure 20.  $t_{Loop1}$  Setup and Timing Diagram

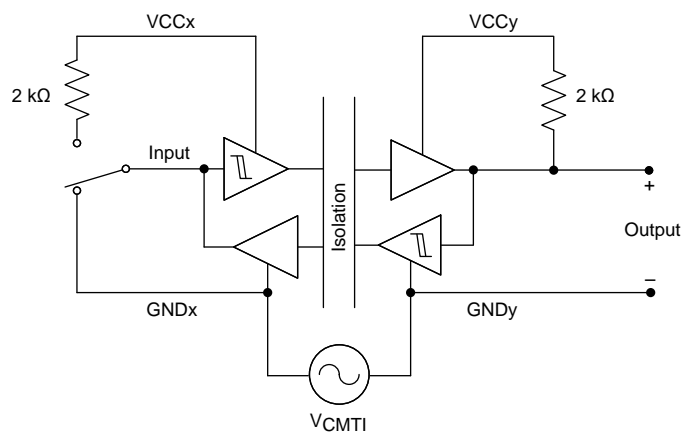
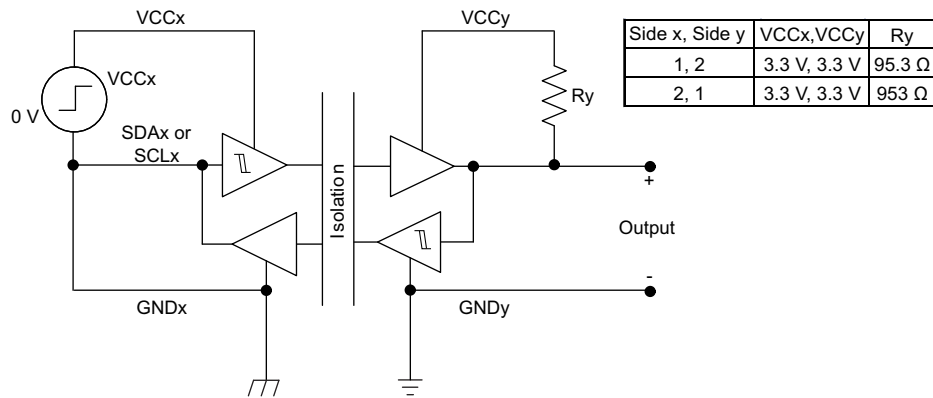


Figure 21. Common-Mode Transient Immunity Test Circuit



Parameter Measurement Information (continued)



or

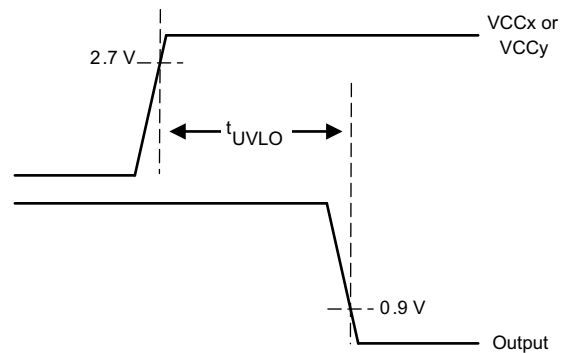
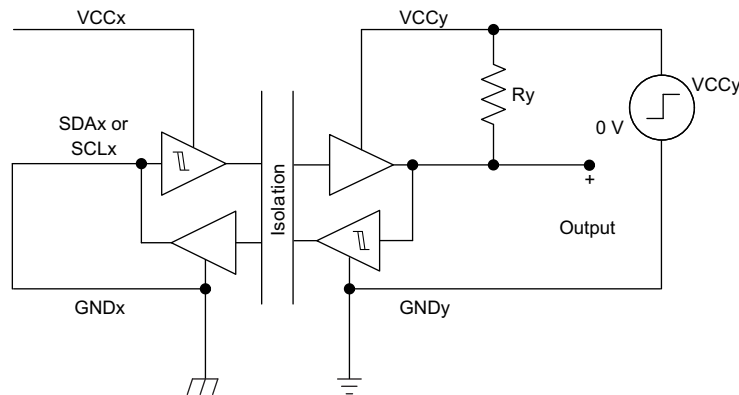


Figure 22.  $t_{UVLO}$  Test Circuit and Timing Diagrams

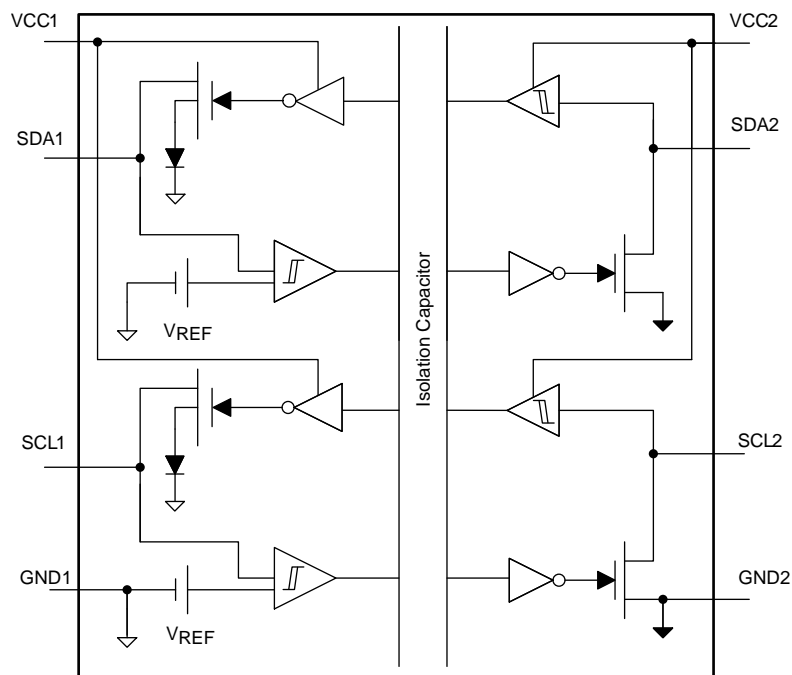
## 8 Detailed Description

### 8.1 Overview

The I<sup>2</sup>C bus is used in a wide range of applications because it is simple to use. The bus consists of a two-wire communication bus that supports bidirectional data transfer between a master device and several slave devices. The master, or processor, controls the bus, specifically the serial clock (SCL) line. Data is transferred between the master and slave through a serial data (SDA) line. This data can be transferred in four speeds: standard mode (0 to 100 kbps), fast mode (0 to 400 kbps), fast-mode plus (0 to 1 Mbps), and high-speed mode (0 to 3.4 Mbps). The most common speeds are the standard and fast modes.

The I<sup>2</sup>C bus operates in bidirectional, half-duplex mode, while standard digital isolators are unidirectional devices. To make efficient use of one technology supporting the other, external circuitry is required that separates the bidirectional bus into two unidirectional signal paths without introducing significant propagation delay. These devices have their logic input and output buffers separated by TI's capacitive isolation technology using a silicon dioxide (SiO<sub>2</sub>) barrier. When used in conjunction with isolated power supplies, these devices block high voltages, isolate grounds, and prevent noise currents from entering the local ground and interfering with or damaging sensitive circuitry.

### 8.2 Functional Block Diagrams



**Figure 23. ISO1540 Block Diagram**

Functional Block Diagrams (continued)

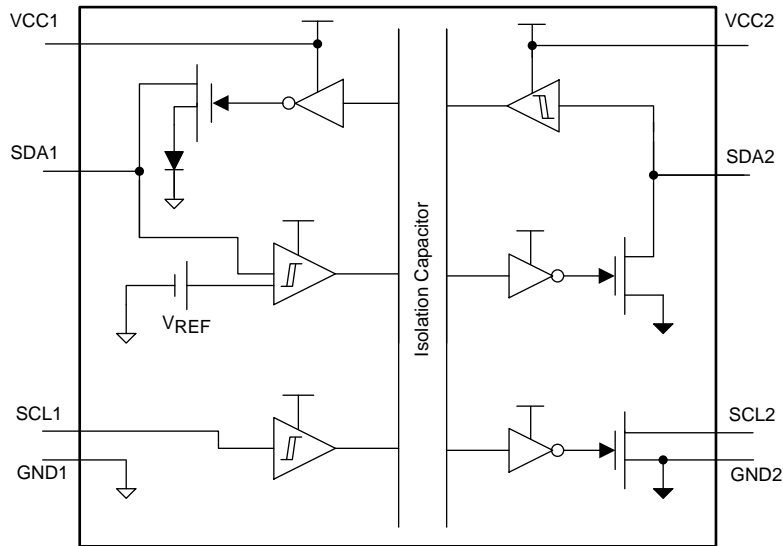


Figure 24. ISO1541 Block Diagram

8.3 Feature Description

The device enables a complete isolated I<sup>2</sup>C interface to be implemented within a small form factor having the features listed in Table 1.

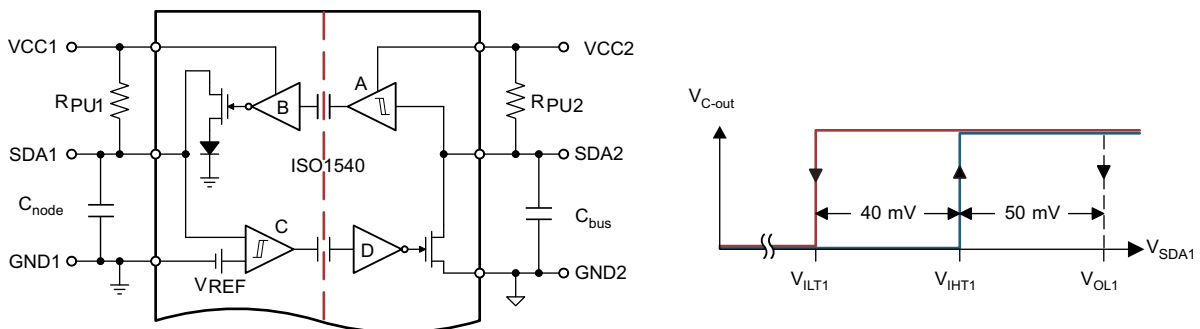
Table 1. Features List

PART NUMBER	CHANNEL DIRECTION	RATED ISOLATION <sup>(1)</sup>	MAXIMUM FREQUENCY
ISO1540	Bidirectional (SCL) Bidirectional (SDA)	2500 V <sub>RMS</sub> 4242 V <sub>PK</sub>	1 MHz
ISO1541	Unidirectional (SCL) Bidirectional (SDA)		

(1) See [Safety-Related Certifications](#) for detailed Isolation specifications.

8.4 Isolator Functional Principle

To isolate a bidirectional signal path (SDA or SCL), the ISO1540 internally splits a bidirectional line into two unidirectional signal lines, each of which is isolated through a single-channel digital isolator. Each channel output is made open-drain to comply with the open-drain technology of I<sup>2</sup>C. Side 1 of the ISO1540 connects to a low-capacitance I<sup>2</sup>C node, while side 2 is designed for connecting to a fully loaded I<sup>2</sup>C bus with up to 400 pF of capacitance.



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Figure 25. SDA Channel Design and Voltage Levels at SDA1

### Isolator Functional Principle (continued)

At first sight, the arrangement of the internal buffers suggests a closed signal loop that is prone to latch-up. However, this loop is broken by implementing an output buffer (B) whose output low-level is raised by a diode drop to approximately 0.75 V, and the input buffer (C) that consists of a comparator with defined hysteresis. The comparator's upper and lower input thresholds then distinguish between the proper low-potential of 0.4 V (maximum) driven directly by SDA1 and the buffered output low-level of B.

Figure 26 demonstrate the switching behavior of the I<sup>2</sup>C isolator, ISO1540, between a master node at SDA1 and a heavy loaded bus at SDA2.

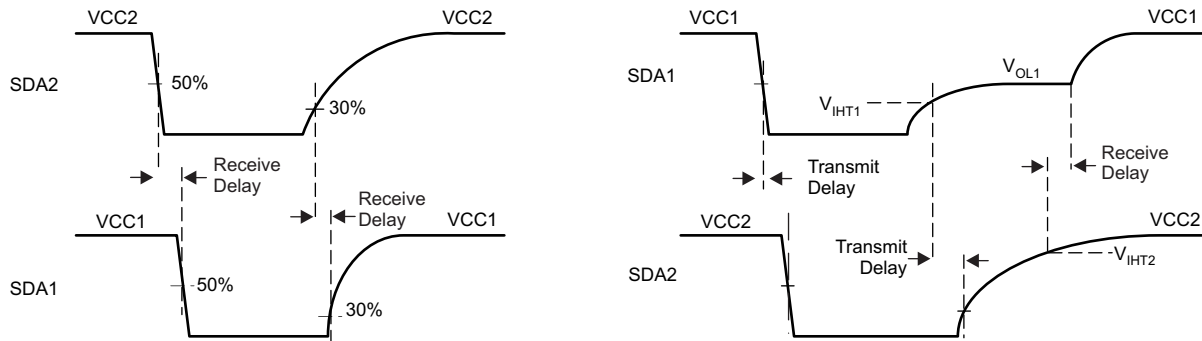


Figure 26. SDA Channel Timing in Receive and Transmit Directions

#### 8.4.1 Receive Direction (Left Diagram of Figure 26)

When the I<sup>2</sup>C bus drives SDA2 low, SDA1 follows after a certain delay in the receive path. The output low is the buffered output of  $V_{OL1} = 0.75\text{ V}$ , which is sufficiently low to be detected by Schmitt-trigger inputs with a minimum input-low voltage of  $V_{IL} = 0.9\text{ V}$  at 3 V supply levels.

When SDA2 is released, its voltage potential increases towards  $V_{CC2}$  following the time-constant formed by  $R_{PU2}$  and  $C_{bus}$ . After the receive delay, SDA1 is released and also rises towards  $V_{CC1}$ , following the time-constant  $R_{PU1} \times C_{node}$ . Because of the significant lower time-constant, SDA1 may reach  $V_{CC1}$  before SDA2 reaches  $V_{CC2}$  potential.

#### 8.4.2 Transmit Direction (Right Diagram of Figure 26)

When a master drives SDA1 low, SDA2 follows after a certain delay in the transmit direction. When SDA2 turns low it also causes the output of buffer B to turn low but at a higher 0.75 V level. This level cannot be observed immediately as it is overwritten by the lower low-level of the master.

However, when the master releases SDA1, the voltage potential increases and first must pass the upper input threshold of the comparator,  $V_{IHT1}$ , to release SDA2. SDA1 then increases further until it reaches the buffered output level of  $V_{OL1} = 0.75\text{ V}$ , maintained by the receive path. When comparator C turns high, SDA2 is released after the delay in transmit direction. It takes another receive delay until B's output turns high and fully releases SDA1 to move toward  $V_{CC1}$  potential.

### 8.5 Device Functional Modes

Table 2 lists the ISO154x functional modes.

Table 2. Function Table<sup>(1)</sup>

POWER STATE	INPUT	OUTPUT
$V_{CC1}$ or $V_{CC2} < 2.1\text{ V}$	X	Z
$V_{CC1}$ and $V_{CC2} > 2.8\text{ V}$	L	L
$V_{CC1}$ and $V_{CC2} > 2.8\text{ V}$	H	Z
$V_{CC1}$ and $V_{CC2} > 2.8\text{ V}$	Z <sup>(2)</sup>	?

(1) H = High Level; L = Low Level; Z = High Impedance or Float; X = Irrelevant; ? = Indeterminate

(2) Invalid input condition as an I<sup>2</sup>C system requires that a pullup resistor to VCC is connected.

## 9 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

#### 9.1.1 I<sup>2</sup>C Bus Overview

The inter-integrated circuit (I<sup>2</sup>C) bus is a single-ended, multi-master, 2-wire bus for efficient inter-IC communication in half-duplex mode.

I<sup>2</sup>C uses open-drain technology, requiring two lines, serial data (SDA) and serial clock (SCL), to be connected to VDD by resistors (see Figure 27). Pulling the line to ground is considered a logic zero while letting the line float is a logic one. This logic is used as a channel access method. Transitions of logic states must occur while the SCL pin is low. Transitions while the SCL pin is high indicate START and STOP conditions. Typical supply voltages are 3.3 V and 5 V, although systems with higher or lower voltages are allowed.

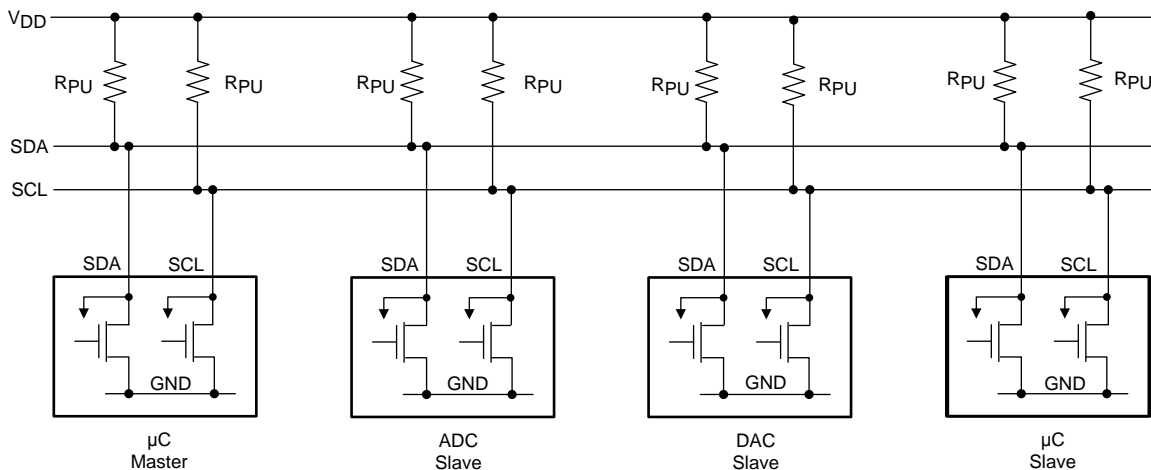


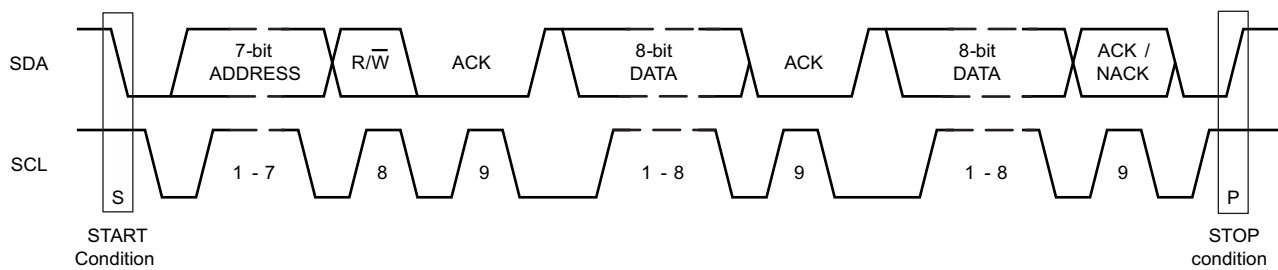
Figure 27. I<sup>2</sup>C Bus

I<sup>2</sup>C communication uses a 7-bit address space with 16 reserved addresses, so a theoretical maximum of 112 nodes can communicate on the same bus. In praxis, however, the number of nodes is limited by the specified, total bus capacitance of 400 pF, which restricts communication distances to a few meters.

The specified signaling rates for the ISO1540 and ISO1541 devices are 100 kbps (standard mode), 400 kbps (fast mode), 1 Mbps (fast mode plus).

The bus has two roles for nodes: master and slave. A master node issues the clock and slave addresses, and also initiates and ends data transactions. A slave node receives the clock and addresses and responds to requests from the master. Figure 28 shows a typical data transfer between master and slave.

**Application Information (continued)**



**Figure 28. Timing Diagram of a Complete Data Transfer**

The master initiates a transaction by creating a START condition, following by the 7-bit address of the slave it wishes to communicate with. This is followed by a single read and write (R/W) bit, representing whether the master wishes to write to 0, or to read from 1 the slave. The master then releases the SDA line to allow the slave to acknowledge the receipt of data.

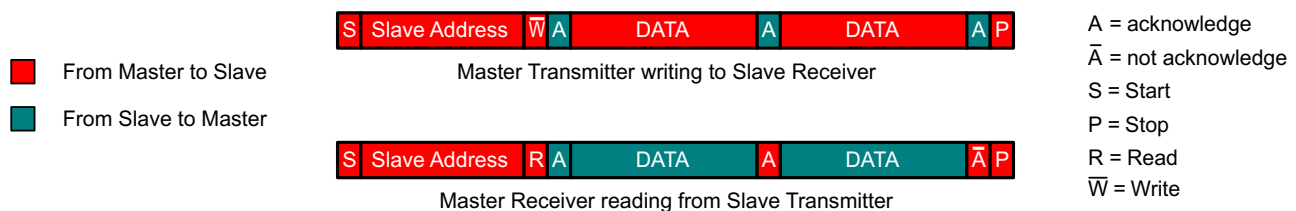
The slave responds with an acknowledge bit (ACK) by pulling the SDA pin low during the entire high time of the 9th clock pulse on the SCL signal, after which the master continues in either transmit or receive mode (according to the R/W bit sent), while the slave continues in the complementary mode (receive or transmit, respectively).

The address and the 8-bit data bytes are sent most significant bit (MSB) first. The START bit is indicated by a high-to-low transition of SDA while SCL is high. The STOP condition is created by a low-to-high transition of SDA while SCL is high.

If the master writes to a slave, it repeatedly sends a byte with the slave sending an ACK bit. In this case, the master is in master-transmit mode and the slave is in slave-receive mode.

If the master reads from a slave, it repeatedly receives a byte from the slave, while acknowledging (ACK) the receipt of every byte but the last one (see Figure 29). In this situation, the master is in master-receive mode and the slave is in slave-transmit mode.

The master ends the transmission with a STOP bit, or may send another START bit to maintain bus control for further transfers.



**Figure 29. Transmit or Receive Mode Changes During a Data Transfer**

When writing to a slave, a master mainly operates in transmit-mode and only changes to receive-mode when receiving acknowledgment from the slave.

When reading from a slave, the master starts in transmit-mode and then changes to receive-mode after sending a READ request (R/W bit = 1) to the slave. The slave continues in the complementary mode until the end of a transaction.

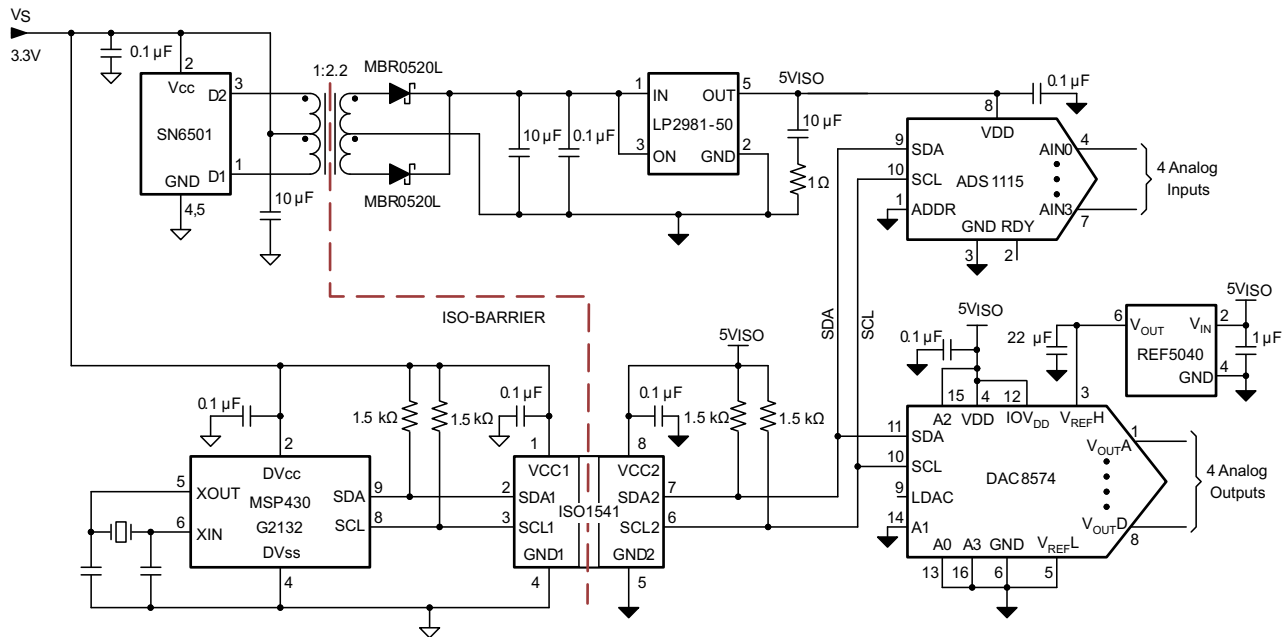
**NOTE**

The master ends a reading sequence by not acknowledging (NACK) the last byte received. This procedure resets the slave state machine and allows the master to send the STOP command.

## 9.2 Typical Application

In Figure 30, the ultra low-power microcontroller, MSP430G2132, controls the I<sup>2</sup>C data traffic of configuration data and conversion results for the analog inputs and outputs. Low-power data converters build the analog interface to sensors and actuators. The ISO1541 device provides the required isolation between different ground potentials of the system controller, remote sensor, and actuator circuitry to prevent ground loop currents that otherwise may falsify the acquired data.

The entire circuit operates from a single 3.3-V supply. A low-power push-pull converter, SN6501, drives a center-tapped transformer with an output that is rectified and linearly regulated to provide a stable 5-V supply for the data converter.



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Figure 30. Isolated I<sup>2</sup>C Data Acquisition System

### 9.2.1 Design Requirements

The recommended power supply voltages (VCC1 and VCC2) must be from 3 V to 5.5 V. A recommended decoupling capacitor with a value of 0.1 µF is required between both the VCC1 and GND1 pins, and the VCC2 and GND2 pins to support of power supply voltages transient and to ensure reliable operation at all data rates.

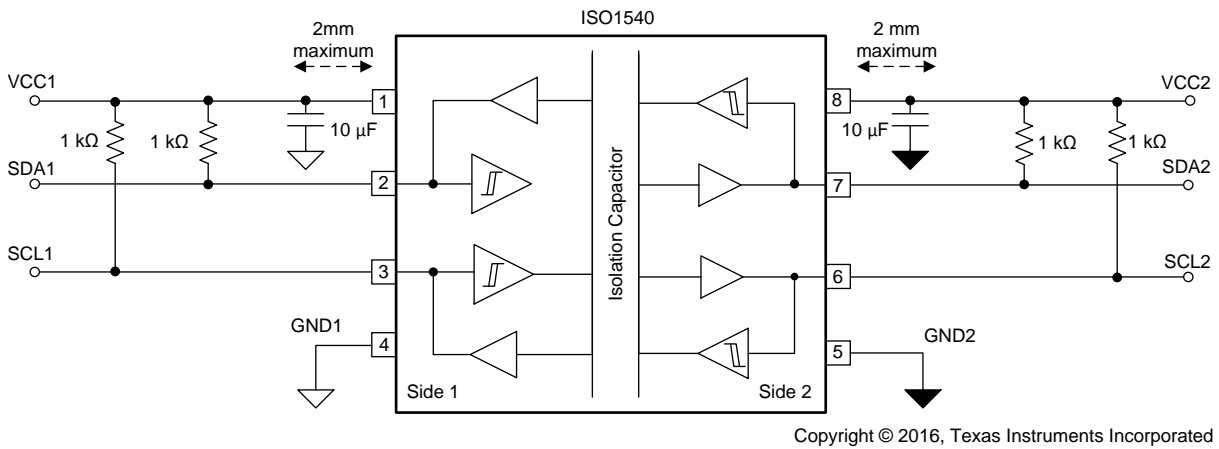
### 9.2.2 Detailed Design Procedure

The power-supply capacitor with a value of 0.1-µF must be placed as close to the power supply pins as possible. The recommended placement of the capacitors must be 2-mm maximum from input and output power supply pins (VCC1 and VCC2).

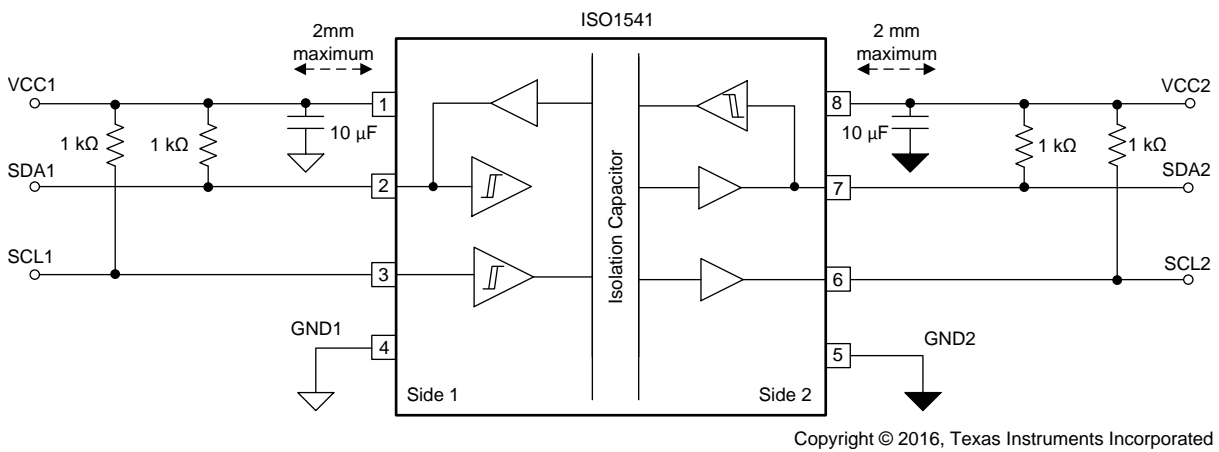
The maximum load permissible on the input lines, SDA1 and SCL1, is  $\leq 40$  pF and on the output lines, SDA2 and SCL2, is  $\leq 400$  pF.

The minimum pullup resistors on the input lines, SDA1 and SCL1 to VCC1 must be selected in such a way that input current drawn is  $\leq 3.5$  mA. The minimum pullup resistors on the input lines, SDA2 and SCL2, to VCC2 must be selected in such a way that output current drawn is  $\leq 35$  mA. The maximum pullup resistors on the input lines (SDA1 and SCL1) to VCC1 and on output lines (SDA1 and SCL1) to VCC2, depends on the load and rise time requirements on the respective lines.

**Typical Application (continued)**

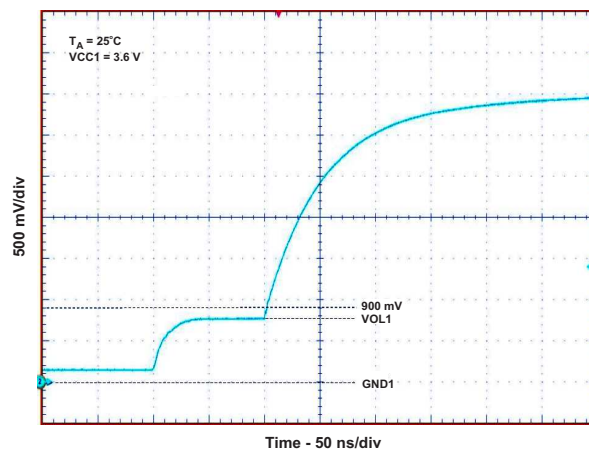


**Figure 31. Typical ISO1540 Circuit Hookup**



**Figure 32. Typical ISO1541 Circuit Hookup**

**9.2.3 Application Curve**



**Figure 33. Side 1: Low-to-High Transition**



## 10 Power Supply Recommendations

To help ensure reliable operation at data rates and supply voltages, TI recommends connecting a 0.1- $\mu$ F bypass capacitor at the input and output supply pins (VCC1 and VCC2). The capacitors should be placed as close to the supply pins as possible. If only a single, primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as TI's [SN6501](#) device. For such applications, detailed power supply design and transformer selection recommendations are available in [SN6501 Transformer Driver for Isolated Power Supplies](#). (SLLSEA0).

## 11 Layout

### 11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see [Figure 34](#)). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

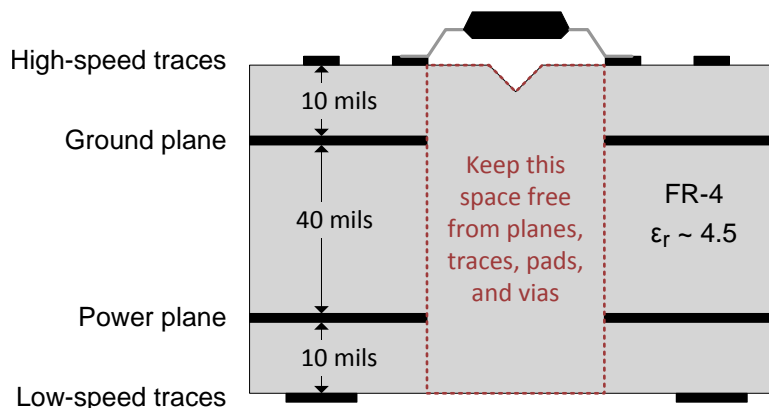
If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, see the [Digital Isolator Design Guide](#) (SLLA284)

#### 11.1.1 PCB Material

For digital circuit boards operating at less than 150 Mbps, (or rise and fall times greater than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit board. This PCB is preferred over cheaper alternatives because of lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and the self-extinguishing flammability-characteristics.

### 11.2 Layout Example



**Figure 34. Recommended Layer Stack**

## 12 Device and Documentation Support

### 12.1 Documentation Support

#### 12.1.1 Related Documentation

For related documentation see the following:

- [Digital Isolator Design Guide](#) (SLLA284)
- [ISO154xEVM Low-Power Bidirectional I<sup>2</sup>C Isolators Evaluation Module](#) (SLLU166)
- [TI Isolation Glossary](#) (SLLA353)
- [SN6501 Transformer Driver for Isolated Power Supplies](#). (SLLSEA0)

### 12.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

**Table 3. Related Links**

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
ISO1540	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
ISO1541	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>

### 12.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.4 Community Resources

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### 12.5 Trademarks

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### 12.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.7 Glossary

**SLYZ022** — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

### **13 Mechanical, Packaging, and Orderable Information**

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
HPA02222DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	IS1540	<a href="#">Samples</a>
ISO1540D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	IS1540	<a href="#">Samples</a>
ISO1540DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	IS1540	<a href="#">Samples</a>
ISO1541D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	IS1541	<a href="#">Samples</a>
ISO1541DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	IS1541	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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**OTHER QUALIFIED VERSIONS OF ISO1540, ISO1541 :**

- Automotive: [ISO1540-Q1](#), [ISO1541-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO1540DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO1541DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO1540DR	SOIC	D	8	2500	367.0	367.0	38.0
ISO1541DR	SOIC	D	8	2500	367.0	367.0	38.0

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
  - E. Reference JEDEC MS-012 variation AA.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



4211283-2/E 08/12

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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